

High Efficiency, 60V Input, 2A Asynchronous Step Down Regulator

General Description

The SY26420 develops a high efficiency, current mode adaptive constant OFF time controlled, asynchronous step-down DC/DC converter capable of delivering 2A output current. The SY26420 operates over a wide input voltage range from 4.5V to 60V and integrates main switch with very low $R_{\rm DS(ON)}$ to minimize the conduction loss. The switching frequency is adjustable from 100kHz to 1MHz using an external resistor. And the device features cycle-by-cycle peak current limitation.

Ordering Information

Ordering Number	Package type	Note
SY26420FCC	SO8E	

Features

- Low $R_{DS(ON)}$ for Internal N-channel Power FET (TOP):175m Ω
- 4.5-60V Input Voltage Range
- 2A Output Current Capability
- Adjustable Switching Frequency Range: 100kHz to1MHz
- Internal Soft-start Limits the Inrush Current
- Hic-cup Mode Output Short Circuit Protection
- EN ON/OFF Control with Accurate Threshold
- Cycle-by-cycle Peak Current Limit
- 0.8V± 1.5% Reference Voltage Accuracy
- Compact Package: SO8E

Applications

- Non-isolated Telecommunication Buck Regulator
- Secondary High Voltage Post Regulator
- Automotive Systems
- Electric Bicycle

Typical Applications

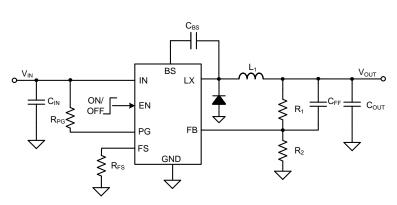


Figure 1. Schematic Diagram

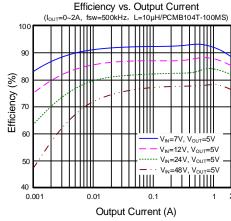
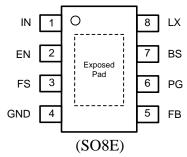


Figure 2. Efficiency vs. Output Current



Pinout (top view)



Top Mark: DUR*xyz* (Device code: DUR; *x=year code*, *y=week code*, *z= lot number code*)

Pin Name	Pin Number	Pin Description
IN	1	Input pin. Decouple this pin to the GND pin with at least a 1µF ceramic capacitor.
EN	2	Enable control. Pulled high to turn on. Do not leave it floating.
FS	3	Frequency programming pin. Connect a resistor to ground to program a switching frequency between 100kHz to 1MHz. The switching frequency equals to: $fsw(kHz)=10^5/R_{FS}(k\Omega)$
GND	4	Ground pin.
FB	5	Output feedback pin. Connect this pin to the center point of the output resistor divider (as shown in Figure 1) to program the output voltage: $V_{OUT}=0.8\times(1+R_1/R_2)$.
PG	6	Power good indicator. Open-drain output when the output voltage is within 90% to 120% of the regulation point.
BS	7	Boot-strap pin. Supply high side gate driver. Connect a 0.1μF ceramic capacitor between the BS pin and the LX pin.
LX	8	Inductor pin. Connect this pin to the switching node of the inductor.
Exposed Pad	/	Exposed pad must be connected to the GND pin. Connected to system ground plane on application board for optimal thermal performance.

Block Diagram

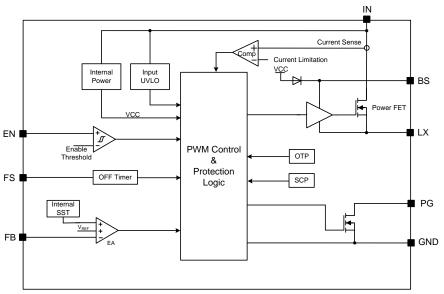


Figure 3. Block Diagram





Absolute Maximum Ratings (Note 1)

8 \ /	
Supply Input Voltage	
BS-LX Voltage	0.3V to 6V
FB, PG, LX Voltage	$-0.3V$ to $V_{IN} + 0.3V$
EN, FS	
Power Dissipation, P _D @ T _A = 25°C, SO8E	2.38W
Package Thermal Resistance (Note 2)	
$\theta_{ m JA}$	42°C/W
$ heta_{ m JC}$	4°C/W
Junction Temperature Range	
Lead Temperature (Soldering, 10 sec.)	260°C
Storage Temperature Range	65°C to 150°C
Dynamic LX voltage in 10 ns Duration	$V_{IN} + 3V$ to GND-5V
Recommended Operating Conditions (Note 3)	
Supply Input Voltage	4.5V to 60V
Junction Temperature Range	40°C to 125°C
Ambient Temperature Range	

Electrical Characteristics

 $(V_{IN}=24V, V_{OUT}=5V, L=10\mu H, C_{OUT}=22\mu F, T_J=-40^{\circ}C$ to $125^{\circ}C, I_{OUT}=0.5A$ unless otherwise specified. The values are guaranteed by test, design or statistical correlation.)

Parameter Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Voltage Range	V_{IN}		4.5		60	V
Input UVLO Threshold	$V_{\rm UVLO}$		3.9	4.2	4.5	V
Input UVLO Hysteresis	V_{HYS}			0.3		V
Quiescent Current	I_Q	V _{FB} =V _{REF} ×105%	70	100	130	μΑ
Shutdown Current	I _{SHDN}	$V_{EN}=0V$	2	6	16	μΑ
Feedback Reference Voltage	V_{REF}		788	800	812	mV
FB Input Current	I_{FB}	$V_{FB}=3.3V$	-50		50	nA
Top FET R _{ON}	R _{DS(ON)}		90	175	265	mΩ
EN Rising Threshold	$V_{\text{EN,R}}$		1	1.1	1.2	V
EN Falling Threshold	$V_{EN,F}$		0.8	0.9	1	V
Min ON Time	t _{ON,MIN}	(Note 4)		150		ns
Min OFF Time	t _{OFF,MIN}	(Note 4)		250		ns
Soft-start Time	t_{SS}	(Note 4)		2		ms
Switching Frequency Program Range	$f_{\rm SW,RNG}$	R _{FS} =100k~1M	100		1000	kHz
Switching Frequency Setting Accuracy	f_{SW}	R _{OSC} =200k	370	500	630	kHz
Power Good Threshold	V_{PG}	V _{FB} falling, PG from high to low	86	90	95	V_{REF}
		V _{FB} rising, PG from low to high	91	92	99	V_{REF}
		V _{FB} rising, PG from high to low	115	120	124	$%V_{REF}$
		V _{FB} falling, PG from low to high	110	115	120	$%V_{REF}$
Down Cood Dolov	t _{PG_F}	High to low (Note 4)		20		μs
Power Good Delay	t _{PG_R}	Low to high (Note 4)		200		μs





Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Power Good Output Low	$V_{PG,L}$	I _{PG} =2mA	0.06	0.1	0.18	V
PG High Leakage Current					1	μΑ
Top FET Current Limit	I _{LMT,RNG}		2.5	3.5	4.7	A
Output Under Voltage Protection Threshold	V _{UVP}	(Note 4)	45	50	55	$%V_{REF}$
Output UVP Delay	t _{UVP,DLY}	(Note 4)		10		μs
UVP Hic-cup ON Time	t _{UVP,ON}	(Note 4)		2		ms
UVP Hic-cup OFF Time	t _{UVP,OFF}	(Note 4)		16		ms
Thermal Shutdown Temperature	T _{SD}	(Note 4)		150		°C
Thermal Shutdown Hysteresis	T _{HYS}	(Note 4)		15		°C

Note 1: Stresses beyond "Absolute Maximum Ratings" may cause permanent damage to the device. These are for stress ratings. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Note 2: θ_{JA} is measured in the natural convection at $T_A = 25^{\circ}C$ on a two-layer Silergy demo board.

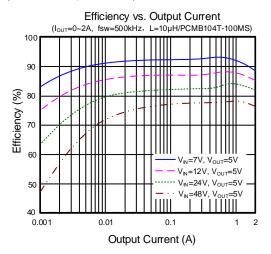
Note 3: The device is not guaranteed to function outside its operating conditions.

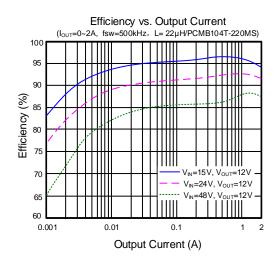
Note 4: Guaranteed by design.

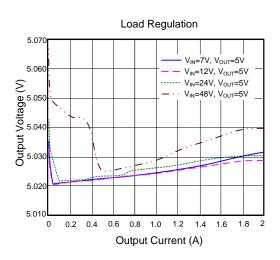


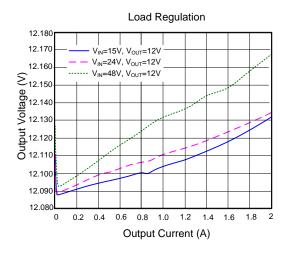
Typical Performance Characteristics

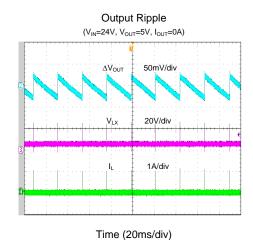
 $(f_{SW}=500kHz, T_A=25^{\circ}C)$

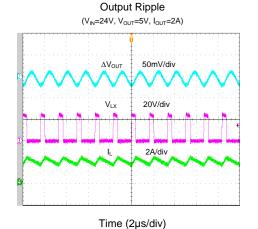




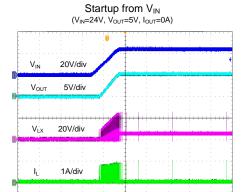




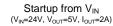


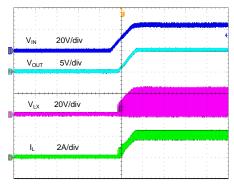






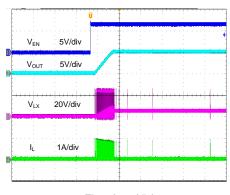
Time (2ms/div)





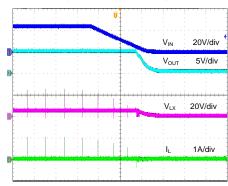
Time (2ms/div)

Startup from Enable (V_{IN}=24V, V_{OUT}=5V, I_{OUT}=0A)



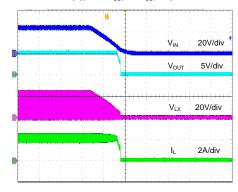
Time (2ms/div)

Shutdown from V_{IN} (V_{IN}=24V, V_{OUT}=5V, I_{OUT}=0A)



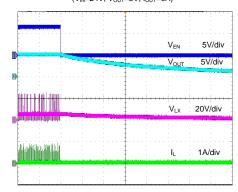
Time (20ms/div)

Shutdown from V_{IN} (V_{IN} =24V, V_{OUT} =5V, I_{OUT} =2A)



Time (20ms/div)

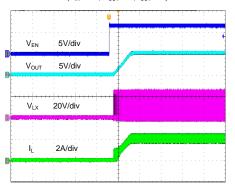
Shutdown from Enable $(V_{IN}\!\!=\!\!24V,\,V_{OUT}\!\!=\!\!5V,\,I_{OUT}\!\!=\!\!0A)$



Time (1s/div)

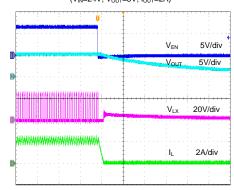


Startup from Enable (V_{IN}=24V, V_{OUT}=5V, I_{OUT}=2A)



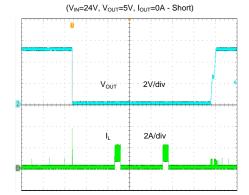
Time (2ms/div)

Shutdown from Enable (V_{IN}=24V, V_{OUT}=5V, I_{OUT}=2A)



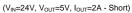
Time (20µs/div)

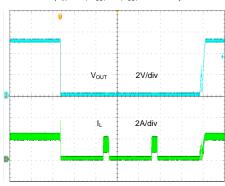
Short Circuit Protection



Time (10ms/div)

Short Circuit Protection

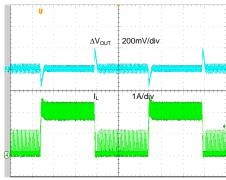




Time (10ms/div)

Load transient





Time (200µs/div)



Operation

The SY26420 is a high efficiency asynchronous step down DC/DC regulator capable of delivering 2A output current. The device adopts current mode adaptive constant OFF time control. The SY26420 operates over a wide input voltage range from 4.5V to 60V and integrates main switch with very low RDS(ON) to minimize the conduction loss.

The switching frequency is adjustable from 100kHz to 1MHz using an external resistor. And the device features cycle-by-cycle peak current limitation.

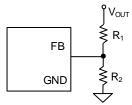
Applications Information

Because of the high integration in the SY26420, the application circuit based on this IC is rather simple. Only the input capacitor $C_{\rm IN}$, the output capacitor $C_{\rm OUT}$, the output inductor L_1 and the feedback resistors (R_1 and R_2) need to be selected for the target applications.

Feedback Resistor Dividers R₁ and R₂:

Choose R_1 and R_2 to program the proper output voltage. To minimize the power consumption under light load, it is desirable to choose large resistance values for both R_1 and R_2 . A value of between $10k\Omega$ and $1M\Omega$ is highly recommended for both resistors. If V_{OUT} is 5V, R_1 = $105k\Omega$ is chosen, then using the following equation, R_2 can be calculated to be $20k\Omega$:

$$R_2 \!=\! \frac{V_{FB}}{V_{OUT} \!-\! V_{FB}} R_1$$
 , V_{FB} is typical 0.8V.



Input Capacitor CIN:

The ripple current through the input capacitor is calculated as:

$$I_{\text{CIN, RMS}} = I_{\text{OUT}} \times \sqrt{D(1-D)}$$

To minimize the potential noise problem, a typical X5R or better grade ceramic capacitor should be placed really close to the IN and the GND pins. Care should be taken to minimize the loop area formed by $C_{\rm IN}$, and the IN/GND pins. In this case, a $1\mu F$ low ESR ceramic capacitor is recommended.

Output Capacitor Cout:

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For the best performance, it is recommended to use an X5R or better grade ceramic capacitor greater than $22\mu F$ capacitance.

Output Inductor L:

There are several considerations in choosing this inductor.

1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum output current. The inductance is calculated as:

$$L = \frac{V_{\text{OUT}}(1 - V_{\text{OUT}}/V_{\text{IN,MAX}})}{f_{\text{SW}} \times I_{\text{OUT,MAX}} \times 40\%}$$

Where fsw is the switching frequency and the I_{OUT MAX} is the maximum load current.

The SY26420 regulator is quite tolerant of different ripple current amplitude. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

 The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

$$I_{\text{SAT, MIN}} > I_{\text{OUT, MAX}} + \frac{V_{\text{OUT}}(1\text{-}V_{\text{OUT}}/V_{\text{IN,MAX}})}{2\times f_{\text{SW}} \times L}$$

3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with DCR<50m Ω to achieve a good overall efficiency.

Enable Operation

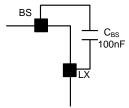
Pulling the EN pin low (<0.8V) will shut down the device. During the shutdown mode, the SY26420 shutdown current will drop to lower than $6\mu A(typ.)$. Driving the EN pin high (>1.2V) will turn on the IC again.

It is not recommended to connect EN and IN directly. A resistor in a range of $10k\Omega$ to $1M\Omega$ should be used if EN is pulled high by IN.



External Bootstrap Capacitor

This capacitor provides the gate driver voltage for internal high side MOSFET. A 100nF low ESR ceramic capacitor connected between the BS pin and the LX pin is recommended.



Adjustable Switching Frequency:

The FS pin is used for setting the switching frequency of the device by connecting a resistor from the FS pin to GND. The switching frequency of the device is adjustable from 100 kHz to 1MHz.

 $fsw(kHz)=10^5/R_{FS}(k\Omega)$

The SY26420 linearly folds back the switching frequency when the FB voltage is within 0~50% of the reference voltage during soft-start and the SCP event. The minimum switching frequency is clamped to 50kHz.

Power Good Indication

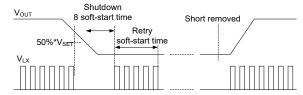
PG is an open-drain output pin. This pin will be pulled to ground if the output voltage is lower than 90% of regulation voltage or higher than 120% of regulation voltage. Otherwise this pin will go to a high impedance state.

Over-current Protection (OCP)

The SY26420 features cycle-by-cycle peak current limit. When the peak current limit is triggered, the device will turn off the high side MOSFET and turn on the low side MOSFET until the next cycle clock is coming.

Short-circuit Protection (SCP)

If $V_{\rm OUT} < 50\%$ of the set-point continuously for approximately $10\mu s$, the short-circuit protection mode will be initiated, and the device will shut down for 8 soft-start time. The device will then restart with a complete soft-start cycle. If the short circuit condition remains, another 'hiccup' cycle of shutdown and restart will continue indefinitely.



Over-temperature Protection (OTP)

The device includes over-temperature protection (OTP) circuitry to prevent overheating due to excessive power dissipation. This will shut down switching operation when the junction temperature exceeds 150°C. Once the junction temperature cools down by approximately 15°C the IC will resume normal operation after a complete soft-start cycle. For continuous operation, provide adequate cooling so that the junction temperature does not exceed the OTP threshold.

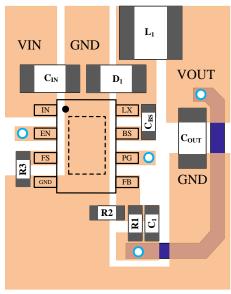
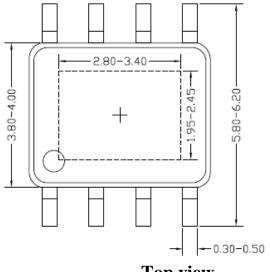
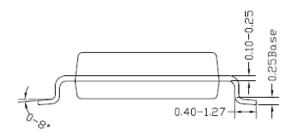


Figure 4. PCB Layout Suggestion

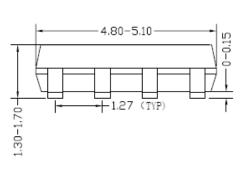


SO8E Package Outline & PCB Layout

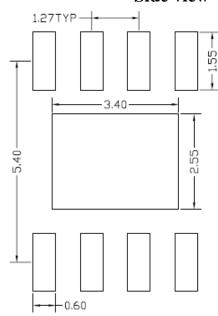




Top view



Side view



Front view

Recommended PCB layout (Reference Only)

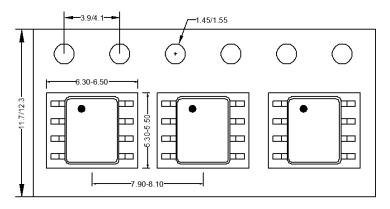
All dimension in millimeter and exclude mold flash & metal burr. **Notes:**



Taping & Reel Specification

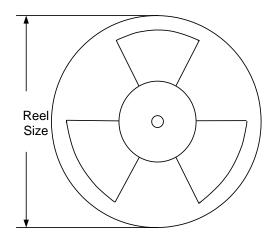
1. Taping orientation

SO8E



Feeding direction -----

2. Carrier Tape & Reel specification for packages



Package types	Tape width	Pocket	Reel size	Reel size		Qty per reel	
i ackage types	(mm)	pitch(mm)	(Inch)	length(mm)	length (mm)	(pcs)	
SO8E	12	8	13"	400	400	2500	

Others: NA



Revision History

The revision history provided is for informational purpose only and is believed to be accurate, however, not warranted. Please make sure that you have the latest revision.

Date	Revision	Change
July 09, 2024	Revision 1.0	Fix a typo in Features:
		Reference Voltage Accuracy is 0.8V±1.5%, (Consistent with EC table)
Apr.16, 2021	Revision 0.9	Initial Release





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