

### General Description

The SY28480 is a 0.76mΩ ultra-low on-resistance E-fuse for using in high current applications such as base stations and servers.

The SY28480 integrates overcurrent protection to against over load and output hard short event. The SY28480 also offer an analog output for high accurate load current monitor and provide programmable soft start time to fit different output cap and reduce the inrush current.

The SY28480 is available in a compact QFN5x5-32 package.

### Applications

- Servers
- Base Station
- Hot Swap Applications

### Features

- Wide Input Range: 4.5 V to 18 V
- Up to 60 A Peak Current Output, 50 A Continuous
- Integrated N-Channel MOSFET with 0.76 mΩ Ultra Low R<sub>ON</sub>
- Adjustable Slew Rate Control
- Adjustable Current Limit
- Accurate Analog Load Current Monitor
- Adjustable Over Current Alert Output
- Temperature Indicator
- Fault Detection with Status OK Output
- Can be used in Parallel for Higher Current Applications
- Auto-Retry Mode for Following Protection Features
  - ◆ Soft Start Duration Timeout
  - ◆ Thermal Shutdown
  - ◆ Fast Short-Circuit Protection
  - ◆ Current Limiting Response Timeout
- Compact Package Minimizes the Board Space: QFN 5 mm × 5 mm-32

### Typical application

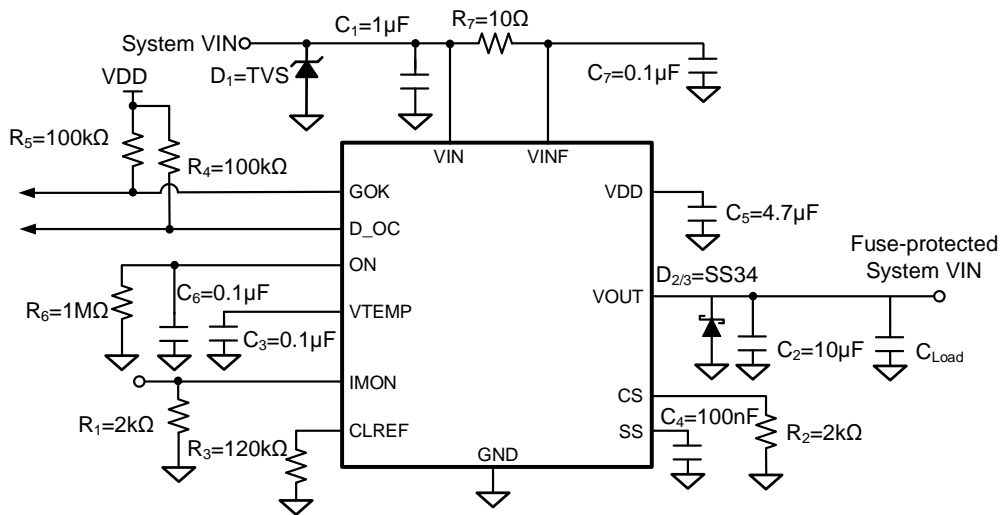


Figure1. Schematic Diagram

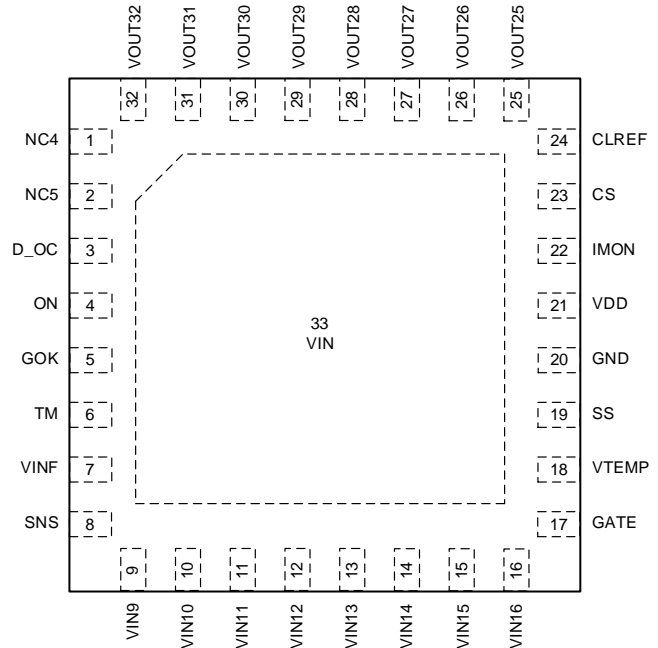
## Ordering Information

Ordering Part Number	Package Type	Top Mark
SY28480QEQ	QFN5x5-32 RoHS Compliant and Halogen Free	DQDxyz

Device code: DQD

*x=year code, y=week code, z= lot number code*

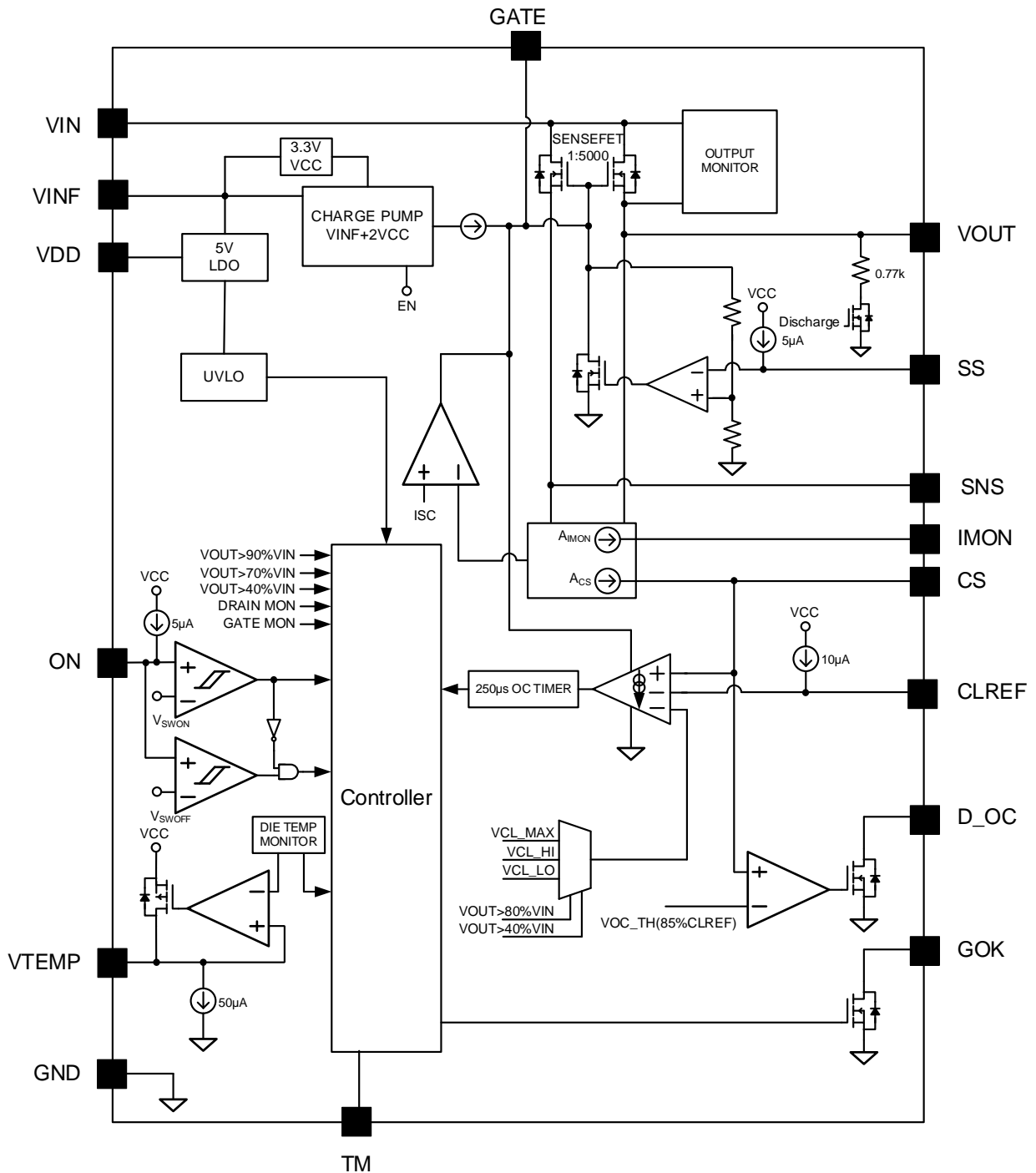
## Pinout (top view)



Pin Name	Pin number	Pin Description
NC4	1	No electrical connection internally. May connect to any potential.
NC5	2	No electrical connection internally. May connect to any potential.
D_OC	3	Over current indicator output (open drain). Low indicates the device is limiting current. The D_OC output does not report current limiting during soft-start.
ON	4	Enable output pull down resistance control.
GOK	5	OK status indicator output (open drain). Low indicates that the device was turned off by a fault.
TM	6	Test pin. Do not connect to this pin. Leave floating.
VINF	7	Control circuit power supply input. Connect to VIN pins through an RC filter.
SNS	8	Internal FET sense pin. Do not connect to this pin. Leave floating.
VIN09	9	Input of high current output switch.
VIN10	10	Input of high current output switch.
VIN11	11	Input of high current output switch.
VIN12	12	Input of high current output switch.
VIN13	13	Input of high current output switch.
VIN14	14	Input of high current output switch.
VIN15	15	Input of high current output switch.
VIN16	16	Input of high current output switch.
GATE	17	Internal FET gate pin. Do not connect to this pin. Leave floating.
VTEMP	18	Analog temperature monitor output.
SS	19	Soft Start time programming pin. Connect a capacitor to this pin to set the soft start time.
GND	20	Ground.

Pin Name	Pin number	Pin Description
VDD	21	Linear regulator output.
IMON	22	Analog current monitor output.
CS	23	Current sense feedback output (current). Scaling the voltage developed at this pin with a resistor to ground makes this also an input for several current limiting functions and over current indicator D_OC.
CLREF	24	Current limit set point input for normal operation (after soft-start).
VOUT25	25	Output of high current output switch.
VOUT26	26	Output of high current output switch.
VOUT27	27	Output of high current output switch.
VOUT28	28	Output of high current output switch.
VOUT29	29	Output of high current output switch.
VOUT30	30	Output of high current output switch.
VOUT31	31	Output of high current output switch.
VOUT32	32	Output of high current output switch.
VIN33	33	Input of high current output switch.

**Block Diagram**



**Figure2. Block Diagram**

## Absolute Maximum Ratings

Parameter (Note 1)	Min	Max	Unit
VINx, VINf	-0.3	20	V
VOUTx	-0.3V/-1V(<500ms)	20	
VDD	-0.3	6	
GOK, D_OC, ON, VTEMP, CLREF, IMON, CS, SS	-0.3	VDD+0.3	
Lead Temperature (Soldering, 10s)		260	°C
Junction Temperature, Operating	-40	150	
Storage Temperature	-65	150	

## Thermal Information

Parameter (Note 2)	Typ	Unit
$\theta_{JA}$ Junction-to-Ambient Thermal Resistance	28	°C/W
$\theta_{JC}$ Junction-to-Case Thermal Resistance	21.3	
$P_D$ Power Dissipation $T_A = 25^\circ\text{C}$	3.6	W

## Recommended Operating Conditions

Parameter (Note 3)	Min	Max	Unit
VIN, VINf	4.5	18	V
VDD, GOK, D_OC, ON, VTEMP, CLREF, IMON, CS, SS	0	5.5	
Maximum Continuous Output Current: $I_{AVE}$		50	A
Peak Output Current: $I_{PEAK}$		60	
VDD Output Load Capacitance Range: $C_{VDD}$	2.2	10	μF
VTEMP Output Load Capacitance Range: $C_{VTEMP}$	0.1		
Soft-start Duration: $t_{SS}$	10	100	ms
CS Load Resistance Range: $R_{CS}$	1.8	4	kΩ
CLREF Voltage Range: $V_{CLREF}$	0.2	1.4	V
Junction Temperature, Operating	-40	125	°C

## Electrical Characteristics

(VINX=VIN=12V, VON=3.3V, CVINF=0.1μF, CVDD=4.7μF, CVTEMP=0.1μF, RVTEMP = 1kΩ, CSS=100nF, TJ=-40°C to 125°C, typical values are TJ=25°C, unless otherwise specified. The values are guaranteed by test, design or statistical correlation)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
<b>VINF INPUT</b>						
Quiescent Current	I <sub>q</sub>	V <sub>ON</sub> >1.4V, null load		1.44	3.0	mA
		V <sub>ON</sub> >1.4V, fault			3.0	mA
		V <sub>ON</sub> <0.8V		1.44	3.0	mA
		V <sub>ON</sub> <0.8V, VINF=16V			3.0	mA
<b>VDD REGULATOR</b>						
VDD Output Voltage	V <sub>DD_NL</sub>	I <sub>VDD</sub> = 0 mA, VINF = 6 V	4.6	4.95	5.2	V
VDD Load Capability	I <sub>DDLOAD</sub>	VINF = 5.5 V			30	mA
VDD Current Limit	I <sub>DD_CL</sub>	VINF = 12 V and VINF = 6 V	40	70		mA
VDD Dropout Voltage		I <sub>VDD</sub> = 25 mA, VINF = 4.5 V		85	200	mV
UVLO Rising Threshold	V <sub>DD_UVR</sub>		4.1	4.3	4.5	V
UVLO Falling Threshold	V <sub>DD_UVF</sub>		3.8	4.0	4.2	V
<b>ON INPUT</b>						
Bias Current	I <sub>ON</sub>	From pin into a 0 V or 1.5 V source	4.0	5.0	6.0	μA
Switch ON Threshold	V <sub>SWON</sub>		1.3	1.4	1.5	V
Switch OFF Threshold	V <sub>SWOFF</sub>			1.2		V
Switch ON Delay Timer	t <sub>ON</sub>	From ON transitioning above V <sub>SWON</sub> to SS start	0.6	1.0	2.5	ms
Switch OFF Delay Time (Note 4)	t <sub>OFF</sub>	From ON transitioning below V <sub>SWOFF</sub> to GATE pulldown		1		μs
ON Current Source Clamp Voltage	V <sub>ON_CLMP</sub>	Max pull-up voltage of current source		3.0		V
Output Pull-down Resistance	R <sub>PD</sub>	V <sub>OUT</sub> = 12 V, PD mode = 1		0.77		kΩ
<b>SS PIN</b>						
Bias Current	I <sub>SS</sub>	From pin into a 0 V or 1 V source	4.62	5.15	5.62	μA
Gain to VOUT	AVSS		9.6	10	10.4	V/V
SS Pull-down Voltage	V <sub>OL_SS</sub>	0.1 mA into pin during ON delay		3		mV
<b>GOK OUTPUT</b>						
Output Low Voltage	V <sub>OL_GOK</sub>	I <sub>GOK</sub> = 1 mA			0.1	V
Off-state Leakage Current	I <sub>LK_GOK</sub>	V <sub>GOK</sub> = 5 V			1.0	μA

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit	
<b>IMON/CS OUTPUT</b>							
IMON or CS Current (single eFuse) Based on 10 $\mu$ A/A+1.5 $\mu$ A	$I_{IMON/ICS}$	$T_J = 0$ to 85 $^{\circ}$ C	$I_{OUT} = 5$ A (Note 4)		51.5		$\mu$ A
			$I_{OUT} = 10$ A (Note 4)		101.5		$\mu$ A
			$I_{OUT} = 25$ A (Note 4)		251.5		$\mu$ A
			$I_{OUT} = 50$ A (Note 4)		501.5		$\mu$ A
Accuracy (single eFuse)		$T_J = 0$ to 85 $^{\circ}$ C	$I_{OUT} = 5$ A (Note 4)	-6		+6	%
			$I_{OUT} = 10$ A (Note 4)	-4		+4	%
			$I_{OUT} = 25$ A (Note 4)	-4		+4	%
			$I_{OUT} = 50$ A (Note 4)	-4		+4	%
Pre-Biased Offset Current Load for Auto-Zero Op-Amp	$I_{AZ\_BIAS}$			1.5		$\mu$ A	
<b>CURRENT LIMIT &amp; CLREF PIN</b>							
Current Limit Voltage	$V_{CL\_TH}$	If $V_{CS} > V_{CL\_TH}$ current limiting regulation occurs via gate	95	98	101	% $V_{CLREF}$	
Current Limit Trigger Threshold During Start-up	$I_{CL\_LOTH}$	$V_{OUT} < 40\%V_{IN}$ , $V_{CLREF} > 0.15$ V, $R_{CS} = 2k\Omega$ (Note 5)	2			A	
Current Limit Clamp Voltage	$V_{CL\_LO}$	$V_{OUT} < 40\%V_{IN}$ , $V_{CLREF} > 0.15$ V	135	152	165	mV	
	$V_{CL\_HI}$	$40\%V_{IN} < V_{OUT} < 80\%V_{IN}$ , $V_{CLREF} > 0.5$ V	480	504	520	mV	
Max Current Limit Reference Voltage	$V_{CL\_MX}$	$V_{OUT} > 80\%V_{IN}$ , $V_{CLREF} > 1.6$ V	1.55	1.6	1.65	V	
Response Time (Note 4)	$t_{CL\_REG}$	$V_{CS} > V_{CLREF}$ until current limiting		100		$\mu$ s	
CLREF Bias Current	$I_{CL}$	From pin into a 1.2 V source	9.6	10	10.4	$\mu$ A	
CLREF Current Source Clamp Voltage	$V_{CL\_CLMP}$	Max pull-up voltage of current source		3.0		V	
FET Turn-off Timer	$t_{CL\_LA}$	Delay between current limit detection and FET turn-off (GOK = 0)		250		$\mu$ s	
<b>D_OC OUTPUT</b>							
Overcurrent Threshold	$V_{OC\_TH}$	If $V_{CS} > V_{OC\_TH}$ D_OC pin pulls low	83	86	90	% $V_{CLREF}$	
Output Low Voltage	$V_{OL\_DOC}$	$I_{DOC} = 1$ mA			0.1	V	
Off-state Leakage Current	$I_{LK\_DOC}$	$V_{DOC} = 5$ V			1.0	$\mu$ A	
Delay (rising) (Note 4)		$V_{CS} < \text{limit}$ until D_OC rising		1.0		$\mu$ s	
Delay (falling) (Note 4)		$V_{CS} > \text{limit}$ until D_OC falling		1.0		$\mu$ s	
<b>SHORT CIRCUIT PROTECTION</b>							
Current Threshold (Note 4)	$I_{SC}$			80		A	
Response Time (Note 4)	$t_{SC}$	From $I_{OUT} > I_{LIMSC}$ until gate pulldown		500		ns	
<b>VTEMP OUTPUT</b>							
Bias Voltage	$V_{VTEMP25}$	At 25 $^{\circ}$ C		550		mV	
Gain (Note 4)	$A_{VTEMP}$	0 $^{\circ}$ C $\leq T_J \leq 125^{\circ}$ C		10		mV/ $^{\circ}$ C	
Load Capability	$R_{VTEMP}$	At 25 $^{\circ}$ C		1		k $\Omega$	
Pulldown Current	$I_{VTEMP}$	At 25 $^{\circ}$ C		50		$\mu$ A	

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
<b>Thermal Shutdown</b>						
Temperature Shutdown (Note 4)	$T_{TSD}$	GOK pulls down		140		°C
Temperature Shutdown Hysteresis (Note 4)	$T_{HYS}$			20		°C
<b>OUTPUT SWITCH(FET)</b>						
On Resistance	$R_{DS(ON)}$	$T_J = 25^{\circ}C$		0.76	1.0	mΩ
Off-state leakage current	$I_{DSOFF}$	$V_{IN} = 16 V, V_{ON} < 1.2 V, T_J = 25^{\circ}C$			1.0	μA
<b>FAULT DETECTION</b>						
$V_{DS}$ Short Threshold	$V_{DS\_TH}$	Startup postponed if $V_{OUT} > V_{DS\_TH}$ at $V_{ON} > V_{SWON}$ transition		90		%VIN
$V_{DS}$ Short OK Threshold	$V_{DS\_OK}$	Startup resumed if $V_{OUT} < V_{DS\_OK}$ any time after postponed		70		%VIN
$V_{GD}$ Short Threshold	$V_{DG\_TH}$	Startup postponed if $V_G > V_{DG\_TH}$ at $V_{ON} > V_{SWON}$ transition		3.1		V
$V_{GD}$ Short OK Threshold	$V_{DG\_OK}$	Startup resumed if $V_G < V_{DG\_OK}$ any time after postponed		3.0		V
$V_G$ Low Threshold (Note 4)	$V_{G\_TH}$	Restart if $V_{GD} < V_{G\_TH}$ after $t_{SSF\_END}$ or $t_{GATE\_FLT}$		3.4		V
$V_{OUT}$ Low Threshold (Note 4)	$V_{OUTL\_TH}$	Restart if $V_{OUT} < V_{OUTL\_TH}$ after $t_{SSF\_END}$		90		%VIN
Gate Fault Timer (Note 4)	$t_{GATE\_FLT}$	Time from $V_{GD} < V_{G\_TH}$ transition after $t_{SSF\_END}$ completed		200		ms
Startup Timer Failsafe (Note 4)	$t_{SSF\_END}$	Time from $V_{ON} > V_{SWON}$ transition, Max programmable soft-start time		200		ms
<b>AUTO-RETRY</b>						
Auto-Retry Delay	$t_{DLY\_RETRY}$	Delay from power-down to retry of startup		1000		ms

**Note 1:** Stresses beyond the “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

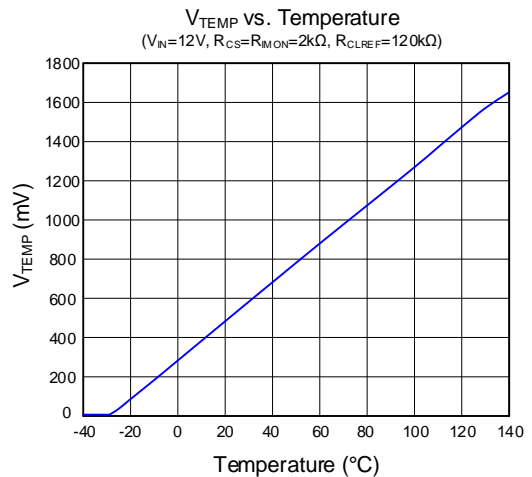
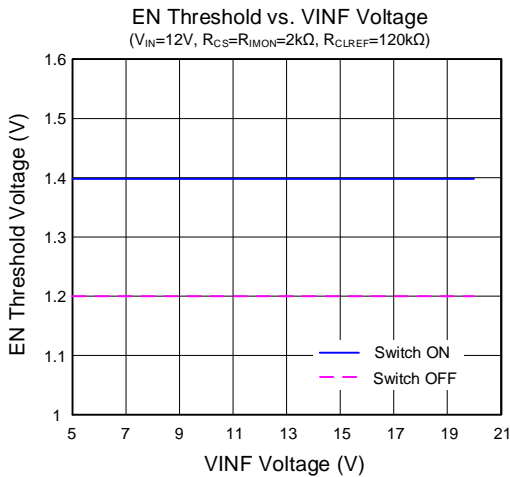
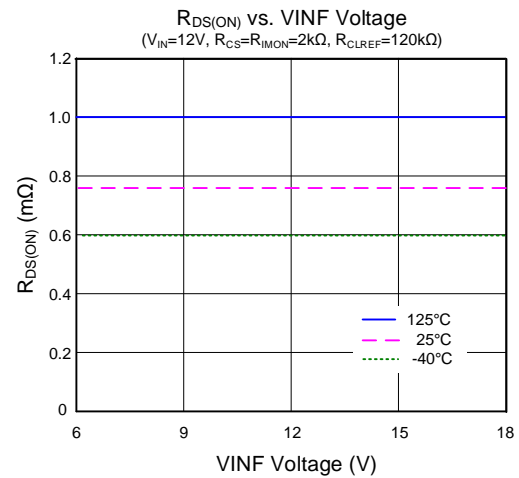
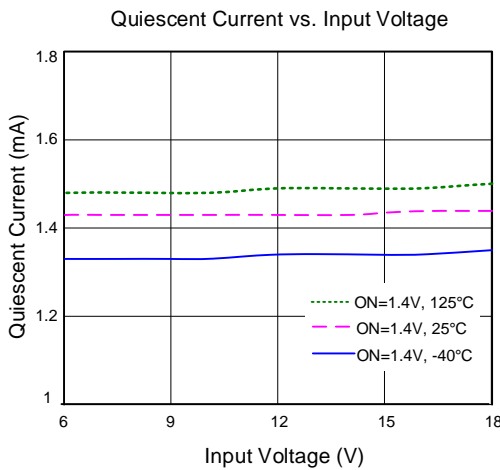
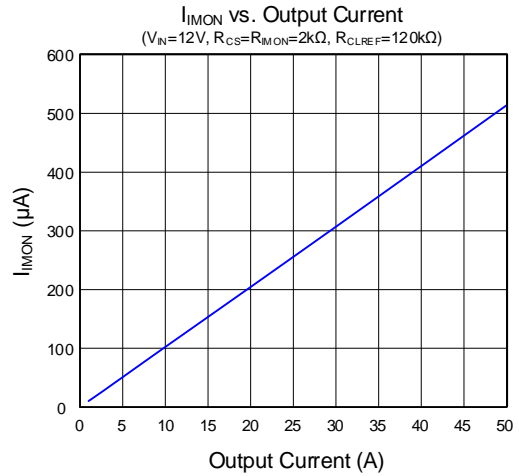
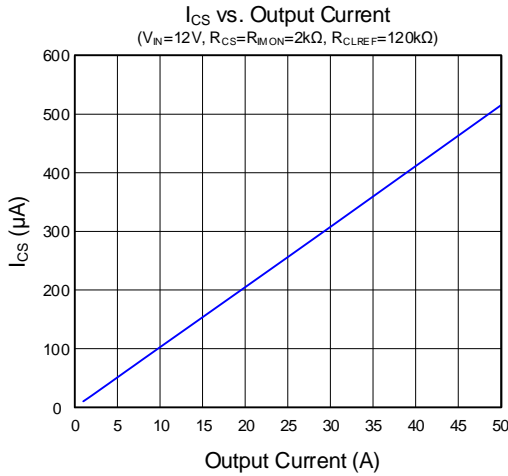
**Note 2:**  $\theta_{JA}$  is simulated in the natural convection at  $T_A = 25^{\circ}C$  on Silergy EVB test board.

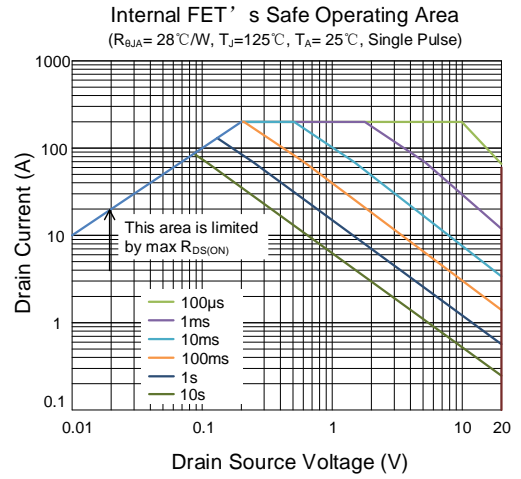
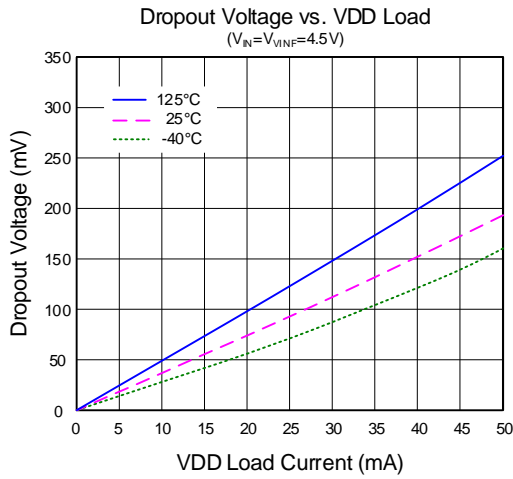
**Note 3:** The device is not guaranteed to function outside its operating conditions.

**Note4:** Guaranteed by design but not production tested.

**Note5:**  $I_{CL\_LOTH}$  indicates the threshold that triggers the timing of current limit when  $V_{OUT} < 40\%V_{IN}$  and  $V_{CLREF} > 0.15V$ , but not the actual current limit clamp threshold. If  $I_{OUT} > I_{CL\_LOTH}$  for a continuous duration  $> t_{CL\_LA}$ , then the device restarts.

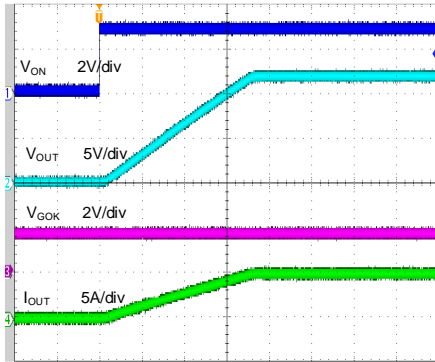
## Typical Performance Characteristics





### Startup from ON

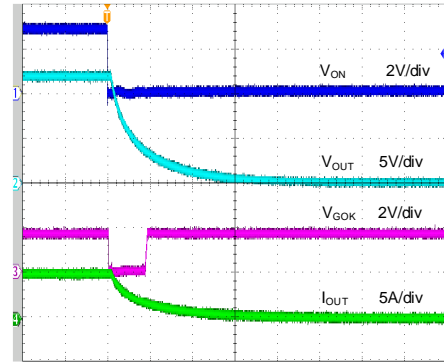
( $V_{IN}=V_{VIN}=12V$ ,  $R_{CS}=R_{MON}=2k\Omega$ ,  $R_{CLREF}=120k\Omega$ ,  $C_{SS}=220nF$ , 2.4 $\Omega$  Load)



Time (20ms/div)

### Shutdown from ON

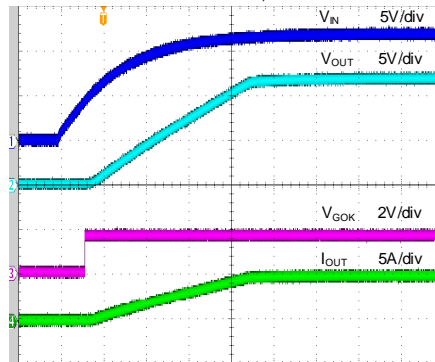
( $V_{IN}=V_{VIN}=12V$ ,  $R_{CS}=R_{MON}=2k\Omega$ ,  $R_{CLREF}=120k\Omega$ ,  $C_{SS}=220nF$ , 2.4 $\Omega$  Load)



Time (20µs/div)

### Startup from $V_{IN}$

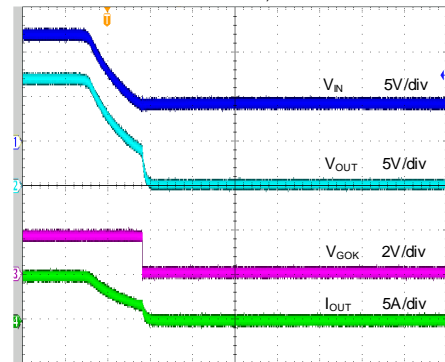
( $V_{IN}=V_{VIN}=12V$ ,  $R_{CS}=R_{MON}=2k\Omega$ ,  $R_{CLREF}=120k\Omega$ ,  $C_{SS}=220nF$ , 2.4 $\Omega$  Load)



Time (20ms/div)

### Shutdown from $V_{IN}$

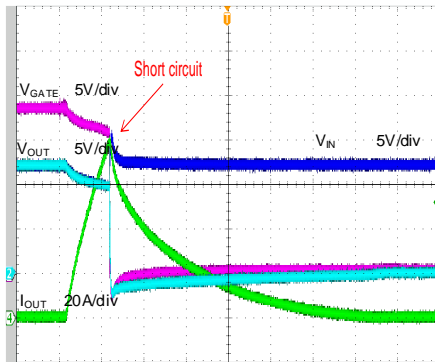
( $V_{IN}=V_{VIN}=12V$ ,  $R_{CS}=R_{MON}=2k\Omega$ ,  $R_{CLREF}=120k\Omega$ ,  $C_{SS}=220nF$ , 2.4 $\Omega$  Load)



Time (400µs/div)

### Short Circuit Response

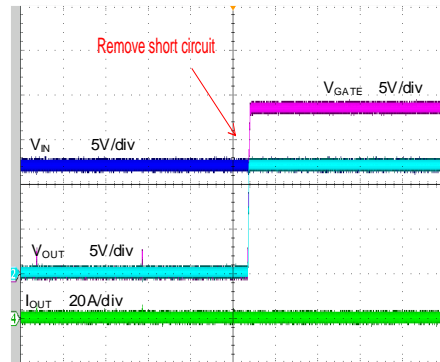
( $V_{IN}=V_{VIN}=12V$ ,  $R_{CS}=R_{MON}=2k\Omega$ ,  $R_{CLREF}=120k\Omega$ ,  $C_{SS}=220nF$ )



Time (40 $\mu$ s/div)

### Short Circuit Recovery

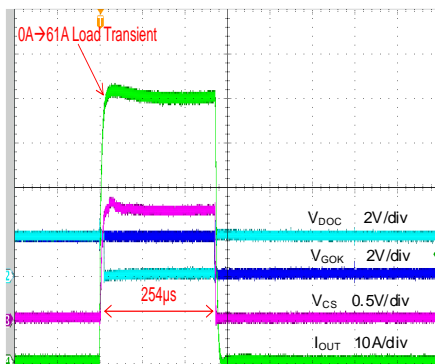
( $V_{IN}=V_{VIN}=12V$ ,  $R_{CS}=R_{MON}=2k\Omega$ ,  $R_{CLREF}=120k\Omega$ ,  $C_{SS}=220nF$ )



Time (400ms/div)

### Current Limit Response During Normal Operation

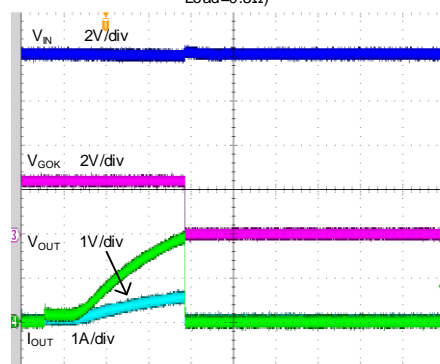
( $V_{IN}=V_{VIN}=12V$ ,  $R_{CS}=R_{MON}=2k\Omega$ ,  $R_{CLREF}=120k\Omega$ ,  $C_{SS}=220nF$ )



Time (100 $\mu$ s/div)

### Current Limit Response During Start-up

( $V_{IN}=V_{VIN}=12V$ ,  $R_{CS}=R_{MON}=2k\Omega$ ,  $R_{CLREF}=120k\Omega$ ,  $C_{SS}=220nF$ , Load=0.3 $\Omega$ )



Time (2ms/div)

## Operation

The SY28480 is a current limited N-channel MOSFET co-packaged with a smart hot swap controller designed for servers or hot swap applications. It incorporates the over temperature protection, over current protection, and short circuit protection. It can be used either alone, or in a parallel configuration for higher current applications.

### VDD Output

The SY28480 provide a 5V output VDD source with current limit upto  $I_{DD\_CL}$ . Shall connect this pin a 2.2 $\mu$ F to 10 $\mu$ F to GND. And place as close as possible to SY28480.

### ON/OFF Control

Without under voltage and any other faults occurring, the output switch will turn on when  $V_{ON}$  is above  $V_{SWON}$ . When  $V_{ON}$  is below  $V_{SWOFF}$ , the switch will turn off.

If  $V_{ON}$  is between  $V_{PD\_OFF}$  and  $V_{SWOFF}$  for longer than  $t_{PD\_DEL}$ , a 0.77k $\Omega$  pull-down resistance to ground is applied to VOUT.

The ON pin sources a 5 $\mu$ A pull up current, connect a capacitor from ON pin to GND can program the startup delay time after  $V_{IN}$  is higher than UVLO.

### Programmable soft start time

The SY28480 can program output start up time by connect a capacitor from SS pin to GND. Soft start capacitor can be calculated by:

$C_{SS} = (t_{SS} \times I_{SS} \times A_{V_{SS}}) / V_{IN}$  (where  $t_{SS}$  is the target soft-start time). The recommended range of  $t_{SS}$  is 10-100ms. The Soft Start Time vs.  $C_{SS}$  curve is shown in Figure3:

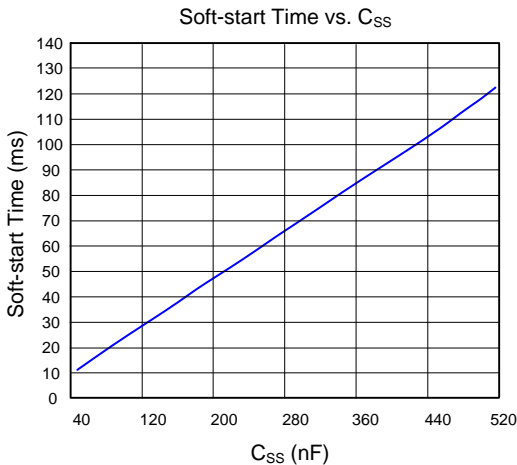


Figure3. Soft Start Time vs.  $C_{SS}$

For parallel application, the SS pins shall be connected together to share one soft start cap so that no slew rate is consistent of each other. The same soft start time make the charging current of each device is same so that the inrush current is balanced.

### GOK Output (Gate OK)

The GOK pin is an open-drain output that is pulled low to report the fault under the following conditions:

VDD	VON	Condition	GOK
<UVLO	X	X	Low
>UVLO	Low	VDS OK fail	Low
>UVLO	Low	VGD OK fail	Low
>UVLO	High	VOUT < 90% * VIN, after soft start done	Low
>UVLO	High	VGS OK fail after $t_{SS\_END}$	Low
>UVLO	High	VGS OK fail after $t_{GATE\_FLT}$	Low
>UVLO	High	Current limit lasts longer than $t_{OC\_LA}$	Low
>UVLO	High	Junction over temperature	Low

Usually, GOK can't be used as power good to indicate the output voltage is in the normal range. Bringing VDD below the UVLO voltage or  $V_{ON} < V_{SWOFF}$  is required to release a latching condition.

### IMON Output (Current Monitor)

The IMON pin sources a current  $A_{IMON}$ , which is proportional (10 $\mu$ A/A) to load current. Connect a resistor from this pin to GND to get voltage signal that can be connect to ADC to monitor the load current. Connect a cap from this pin to GND can act as a low pass filter for ADC converter.

### CLREF Pin (Current Limit and Over-Current Reference)

The voltage on CLREF pin act as a reference of current limit regulation point. This pin internal will source a 10 $\mu$ A current. Connect a resistor from this pin to GND, can program the current limit threshold. The  $V_{CLREF}$  value can be calculated by:  $V_{CLREF} = I_{CL} \times R_{CLREF}$ . The voltage is also can be given by an external voltage source or a DAC. The recommended voltage range for CLREF is between 0.2V to 1.4V.

### CS Input/Output (Current Set)

The CS pin sources a current  $A_{CS}$  that is proportional ( $A_{CS} = 10\mu$ A/A) to load current. Connect a resistor from

this pin to GND to get voltage which act as a feedback signal of current limit regulation loop.

During normal operation ( $V_{ON} > V_{SWON}$  for longer than  $t_{SS\_END}$ ). When the voltage on CS pin, VCS, higher  $V_{CL\_TH}$  ( $V_{CL\_TH}$  is equals to  $V_{CLREF}$ , and will be clamped at  $V_{CL\_MX}$ , if  $V_{CL\_TH} > V_{CL\_MX}$ ), the internal current limit loop will regulate the output current based on formula:  $I_{OUT} = V_{CL\_TH} / (R_{CS} \times A_{CS})$

### **Start up with fold back current**

During startup ( $V_{ON} > V_{SWON}$  for less than  $t_{SS\_END}$ ), the voltage different between VIN and VOUT is large. A folded back current limit threshold can reduce the power dissipation during start up. The fold back current is depend on the VOUT.

- When  $V_{OUT} < 40\%$  of VIN,  $V_{CL\_TH} = V_{CL\_LO}$  or  $V_{CLREF}$  (whichever is lower).
- When VOUT is between 40% and 80% of VIN,  $V_{CL\_TH} = V_{CL\_HI}$  or  $V_{CLREF}$  (whichever is lower).
- When VOUT exceeds 80% of VIN,  $V_{CL\_TH} = V_{CL\_MX}$  or  $V_{CLREF}$  (whichever is lower).

If a current limiting condition exists anytime for a continuous duration  $> t_{CL\_LA}$ , then the device shutdown and restart.

The recommended range of RCS is between 1.8k $\Omega$  and 4k $\Omega$ . This pin do not connect a capacitor from this pin to GND.

### **VINF Pin**

VINF is the power supply input for internal control circuit. Connect an RC filter, a 10 $\Omega$  resistor, and a ceramic capacitor  $\geq 0.1\mu F$  as close as possible to this pin.

### **CS AMP OFFSET BIAS**

The SY28480 pre-biased an offset to keep internal high accurate auto-zero amplifier works at null load or light load condition.

The internal IMON and CS current source follow below relationship:  $I_{OUT} = (I_{CS} - I_{AZ\_BIAS})/10\mu$  and  $I_{OUT} = (I_{MON} - I_{AZ\_BIAS})/10\mu$ .

For typical 1.5 $\mu A$   $I_{AZ\_BIAS}$ , there has 0.15A positive offset in  $I_{OUT}$  sense.

### **DOC Output (Over-current Indicator)**

The D\_OC pin is an open-drain output that indicates when an over current condition exists after soft start is complete. When the voltage on the CS pin, VCS is higher than  $V_{OC\_TH}$ , D\_OC is pulled low. If VCS is lower than  $V_{OC\_TH}$ , D\_OC is high impedance and can be pulled high by external pull up resistor

### **VTEMP Output (Temperature Indicator)**

VTEMP is a voltage output proportional 10mV/ $^{\circ}C$  to device junction temperature, with an offset voltage. if multiple VTEMP outputs are connected, the voltage of all VTEMP outputs will be driven to the voltage produced by the hottest SY28480. A 0.1 $\mu F$  capacitor or greater must be connected from the VTEMP pin to ground. The  $V_{VTEMP}$  value can be calculated by:

$$V_{VTEMP} = V_{VTEMP25} + (T_{TEMP} - 25^{\circ}C) \times A_{VTEMP}$$

(where  $T_{TEMP}$  is the current temperature)

### **Auto-Retry Restart**

When the following protection is triggered, the power FET will be shut down and after a delay time,  $t_{DLY\_RETRY}$ , SY28480 will restart.

Excessive Current Limiting: If a current limiting condition exists anytime for a continuous duration  $> t_{CL\_LA}$ .

Excessive Soft Start Duration: If  $V_{OUT} < V_{OUTL\_TH}$  when  $t_{SSF\_END}$  expires.

Short Circuit Protection: If switch current exceeds  $I_{SC}$   
Over-Temperature Shutdown: If the FET controller temperature  $> T_{TSD}$ .

### **FET Fault Detection**

The device contains various FET monitoring circuits:

- VIN to VOUT short, non-auto-retry condition. If the device is disabled and  $V_{OUT} > V_{DS\_TH}$  then GOK is pulled low and the device is prevented from powering up. The device is allowed to power up once  $V_{OUT} < V_{DS\_OK}$ .
- GATE to VIN short, non-auto-retry condition. If the device is disabled and GATE (Pin 8)  $> V_{DG\_TH}$ , then GOK is pulled low and device is prevented from powering up. The device allowed to power up once  $GATE < V_{DG\_OK}$ .
- GATE leakage-startup.  
If  $(GATE - VINF) < V_{G\_TH}$  at  $t_{SSF\_END}$ , then GOK is pulled low and FET restarts.
- GATE leakage-normal operation.  
If  $(GATE - VINF) < V_{G\_TH}$  for  $t_{GATE\_FLT}$  time after the soft start timer completes, then GOK is pulled low and device restarts.

### **FET SOA Limits**

Power FET's SOA should not be exceeded. In-built timed current limits and fault-monitoring circuit make the power FET work within SOA limits in normal operation status.

### **Multiple Fuse Power Up**

When multiple SY28480 are paralleled together, the SY28480s will turn on together. Keeping the current

through each switch within 1 A (typical) helps to prevent overstress on each switching during soft start.

If all parallel SY28480 SS pins are short circuit, all SY28480s are shutdown when one of them is in Auto-Retry stage because SS is forced to pull down during Auto-Retry stage. Please set all ON pins pull down or reset their VDD and reset circuit to avoid paralleled multiple SY28480 encounter fault.

### Input Filter Capacitor

A 1 $\mu$ F or larger input ceramic capacitor is strongly recommended to be placed close to the VIN. A VOUT short will cause ringing on the VIN without the VIN capacitor. It could cause the N-MOSFET electrical over stress when the VIN transient exceeds the absolute maximum voltage rating even for a short duration.

### Transient Voltage Suppressor

The SY28480 specifies an absolute maximum voltage of 20V and is designed to tolerate brief over voltage events due to hot plug surges. To protect the SY28480 from an over voltage event, install a unidirectional transient voltage suppressor (TVS) such as an SMAJ20CA between the VIN and GND pins.

### OUT Schottky Diode

An anti-parallel schottky diode is suggested to be placed near the VOUT pin to absorb the negative ringing. A low forward voltage and large forward

current schottky diode is recommend.

### Hot Plug Application

In hot plug scenario, the rapid slew rate of input voltage will be coupled to the internal control circuit through the VIN<sub>F</sub>, resulting in internal logic abnormalities. So the VIN<sub>F</sub> needs to be connected to the VIN through an RC filter. Connect an RC filter, a 10 $\Omega$  resistor, and a ceramic capacitor  $\geq 0.1\mu$ F, to reduce the voltage slew rate on VIN<sub>F</sub>.

### PCB Layout Guide

For best performance of the SY28480, the following guidelines must be strictly followed:

1. Keep all power traces as short and wide as possible and use at least 2-ounce copper for all power traces.
2. Place a ground plane under all circuitry to lower both resistance and inductance and improve DC and transient performance.
3. Locate the output capacitor as close to the connectors as possible to lower impedance (mainly inductance) between the port and the capacitor and improve transient performance.
4. Input and output filter capacitors should be placed closed to the IC and connected to ground plane to reduce noise coupling.
5. Locate the RC filter as close to VIN<sub>F</sub> as possible to stabilize the bias supply.

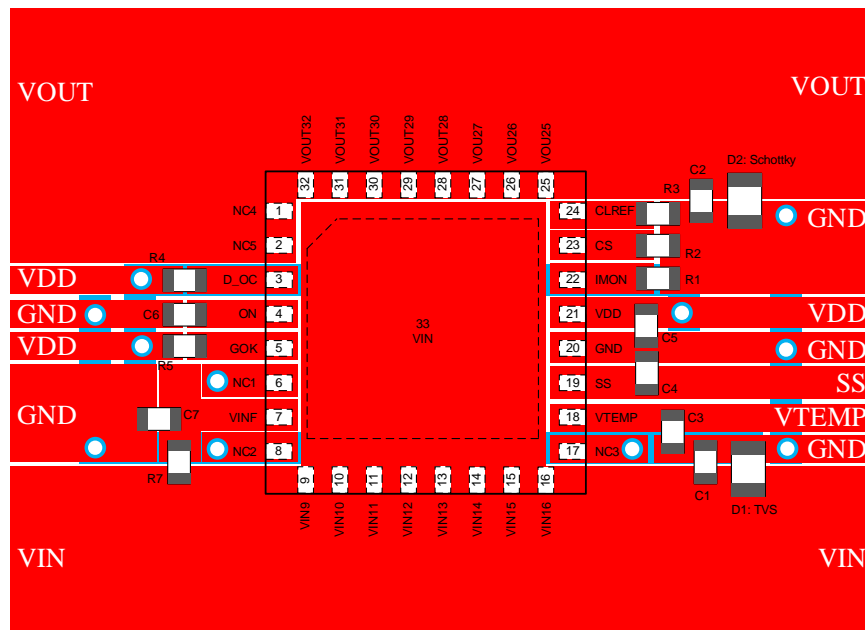
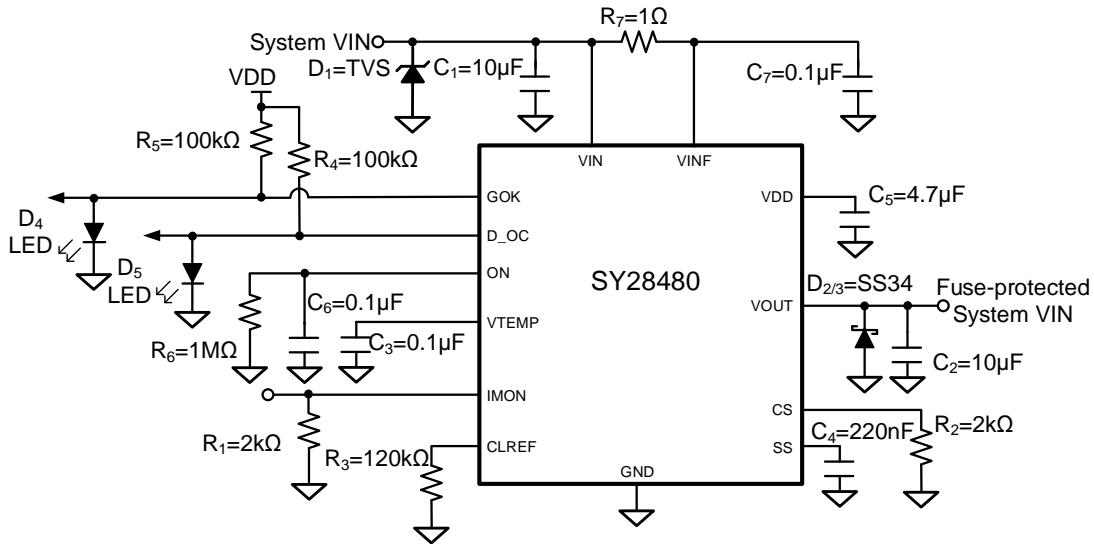


Figure 4. PCB Layout Suggestion

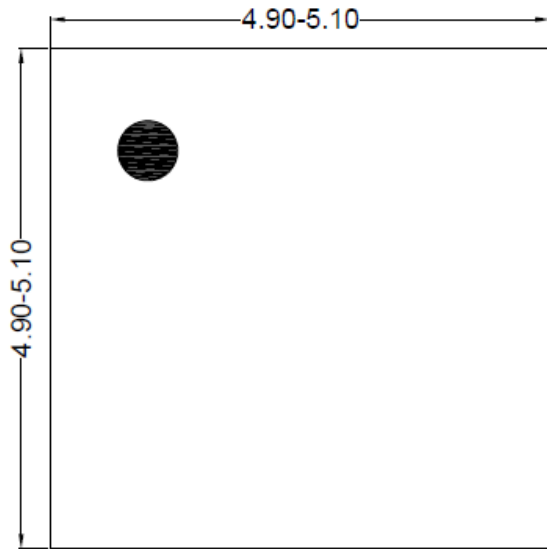
## Schematic



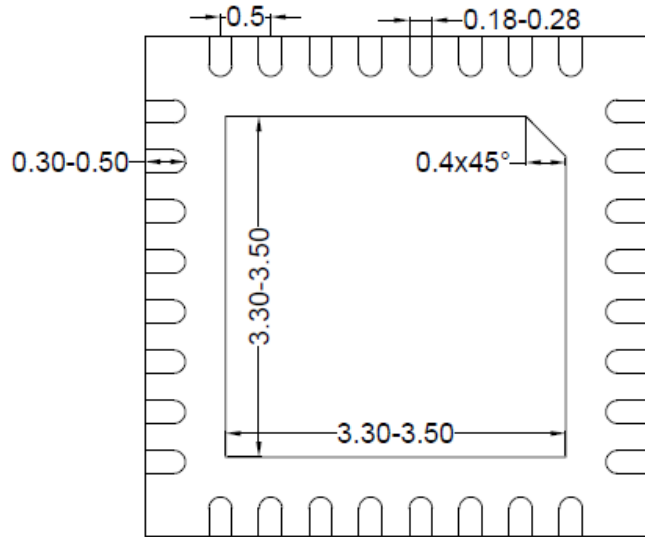
## BOM List

Reference Designator	Description	Part Number	Manufacturer
C1	10μF/50V, 1206	GRT31CR61H106KE01L	Murata
C2	10μF/50V, 1206	GRT31CR61H106KE01L	Murata
C3	0.1μF/50V, 0603	GCJ188R71H104KA12	Murata
C4	220nF/50V, 0603	cGA3E3X7R1H224K080AE	Murata
C5	4.7μF/16V, 0603	GRM185R61C475KE11D	Murata
C6, C7	0.1μF/50V, 0603	GCJ188R71H104KA12	Murata
R1	2kΩ, 0603	RC0603FR-072KL	YAGEO
R2	2kΩ, 0603	RC0603FR-072KL	YAGEO
R3	120kΩ, 0603	RC0603FR-07120KL	YAGEO
R4	100kΩ, 0603	RC0603FR-07100KL	YAGEO
R5	100kΩ, 0603	RC0603FR-07100KL	YAGEO
R6	1MΩ, 0603	RC0603FR-071ML	YAGEO
R7	1Ω, 0603	RC0603FR-071RL	YAGEO
D1	TVS/20V	SMAJ20CA	PHY
D2/3	Schottky/40V	SS34	TOSHIBA
D4/5	LED	/	/

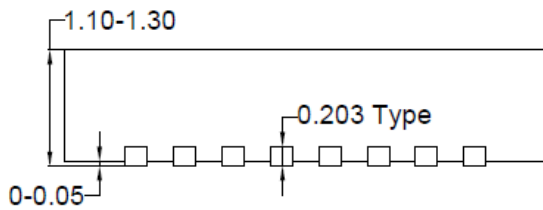
**QFN5.0×5.0-32 Package Outline Drawing**



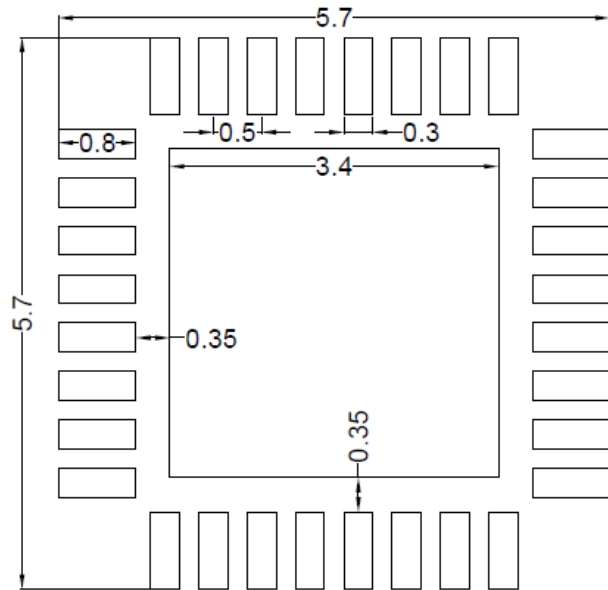
**Top view**



**Bottom view**



**Side view**

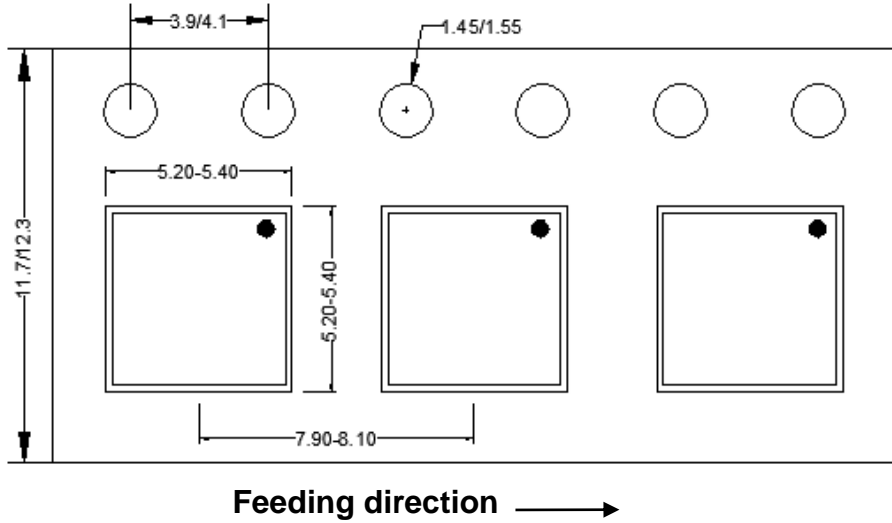


**Recommended PCB layout  
(Reference only)**

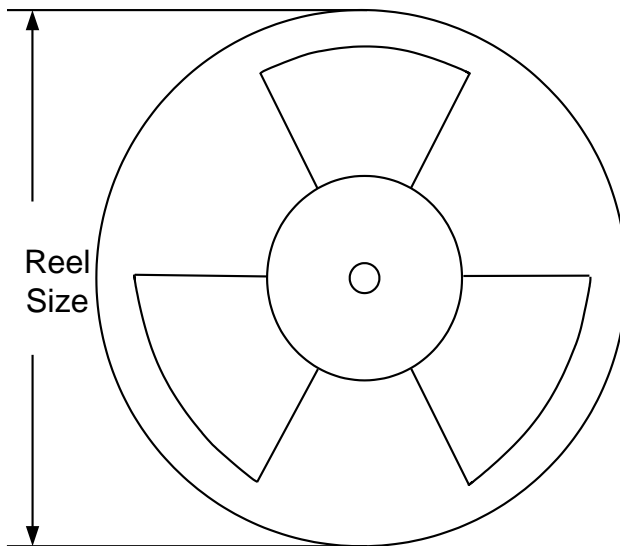
**Notes: 1, All dimension in millimeter and exclude mold flash & metal burr;**

## Taping & Reel Specification

### 1. Taping orientation



### 2. Carrier Tape & Reel specification for packages



Package types	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Trailer length(mm)	Leader length (mm)	Qty per reel
QFN5x5	12	8	13"	400	400	5000

### 3. Others: NA

## Revision History

The revision history provided is for informational purpose only and is believed to be accurate, however, not warranted. Please make sure that you have the latest revision.

Date	Revision	Change
Dec.27, 2025	Revision 1.0A	<ol style="list-style-type: none"> <li>1. The capacitor between the VIN<sub>F</sub> pin and the GND changed from 1μF to 0.1μF;</li> <li>2. Update the application description of “ON/OFF Control ” (Page 12)</li> <li>3. Update the application description of “GOK Output (Gate OK)” (Page 12)</li> <li>4. Update the application description of “VIN<sub>F</sub> Pin” (Page 13)</li> <li>5. Update the application description of “Hot Plug Application” (Page14)</li> </ol>
Dec. 13, 2023	Revision 1.0	Initial Release

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