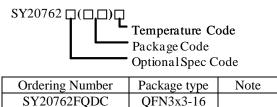


## **General Description**

SY20762F is a 3.0-5.5V<sub>IN</sub>, 2A two-cell synchronous boost Li-Ion battery charger integrates 1MHz switching frequency and full protection functions. The charge current up to 2A can be programmed by using the external resistor for different portable applications and indicates the charger current information simultaneous. It has adaptive input current limit with selectable threshold for safety battery charge operation. SY20762F can disconnect output when there is output short circuit or shutdown happens. It consists of 18V rating FETs with extremely low ON resistance to achieve high charge efficiency and simple peripheral circuit design.

SY20762F along with small QFN3x3 footprint provides small PCB area application.

# **Ordering Information**



## Features

- Low Profile QFN3x3 Package
- Integrated Synchronous Boost with 18V Rating Low R<sub>DSON</sub> FETs for High Charge Efficiency
- Trickle Current / Constant Current / Constant Voltage Charge Mode
- Adaptive Input Current Limit with selectable threshold
- Maximum 2A Constant Charge Current
- Charge Current Information Indication
- Programmable Constant Charge Current
- Programmable Charge Termination Current
- Thermal Regulation Protection
- External Shutdown Function
- Input Voltage UVLO and OVP
- Output Short Circuit Protection
- Over Temperature Protection
- Input Power Good Indication
- Battery Inserted Indication
- Charge Status Indication
- Normal Synchronous Boost Operation When Battery Removed

## **Applications**

- POS, Cellular Telephones, PDA, MP3 Players, MP4 Players
- Digital Cameras
- Bluetooth Applications
- PSP Game Players, NDS Game Players

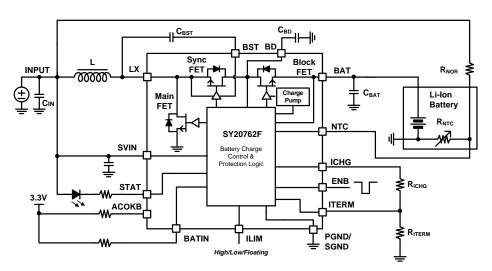
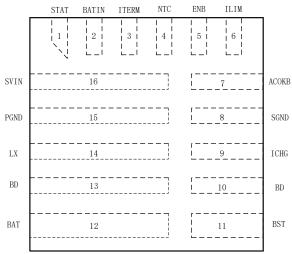


Figure1. Schematic Diagram

## **Typical Applications**



**Pinout (top view)** 



(QFN3x3-16)

Top Mark: AVXxyz, (1	Device code: AVX, <i>x=year code</i> ,	<i>y=week code, z= lot number code</i> )
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Name	Pin Number	Description		
STAT	1	Charge status indication pin. It is open drain output pin and pulls high to SVIN thru a LED to indicate the charge in process. When the charge is done, LED is off.		
BATIN	2	attery inserted indication. Open drain output. Only can be pulled up to a lower than 6V power rail. Pull low if the battery is removed.		
ITERM	3	Charge termination current program pin. I <sub>TERM</sub> =0.1*I <sub>CC</sub> *(1+R <sub>ICHG</sub> /R <sub>ITERM</sub> )		
NTC	4	Thermal protection pin. UTP threshold is typical $75\% V_{SVIN}$ and OTP threshold is typical $25\% V_{SVIN}$ . Pull up to SVIN can disable charge logic and make the IC operate as normal boost regulator. Pull down to ground can shut down the IC.		
ENB	5	Enable control pin. High logic to disable charger, and low logic to enable charger.		
ILIM	6	Adaptive input current limit setting pin. Select the permitted maximum input voltage drop to trigger the input current limit function. Pull high for 500mV voltage drop, pull low for 375mV, floating for 250mV.		
ACOKB	7	Input power good indication. Open drain output. Pull up to SVIN thru a LED to indicate if input is OK. Pull low if the adapter or USB input is present.		
SGND	8	Signal ground pin.		
ICHG	9	Charge current program pin. Pull down to GND with a resistor $R_{ICHG}$ . The mirror current about 1/10000 of the blocking FET current will dump into the external resistor thru ICHG pin and compared to the internal reverence 1V. So $I_{CC}=[1V/(R_{ICHG}+R_{ITERM})]x10000$ , $I_{TC}=[1V/R_{ICHG}+R_{ITERM}]x1000$ .		
BD	10,13	Connect to the Drain of internal Blocking FET. Bypass at least 4.7uF ceramic cap to GND.		
BST	11	Boot-Strap pin. Supply Rectified FET's gate driver. Decouple this pin to LX with 0.1uF ceramic cap.		
BAT	12	Battery positive pin.		
LX	14	Switch node pin. Connect to external inductor.		
PGND	15	Power ground pin.		
SVIN	16	Analog power input pin. Connect a MLCC from this pin to ground to decouple high harmonic noise. This pin has OVP and UVLO function to make the charger operate within safe input voltage area.		



## **Absolute Maximum Ratings**

SVIN, BAT, LX, NTC, STAT, ACOKB, BD, ENB, ICHG, ILIM	18V
BATIN, ITERM	4V
BST-LX Voltage	4V
LX Pin current continuous	5A
Power Dissipation, PD @ TA = 25°C, QFN3X3	2.6W
Package Thermal Resistance	
θ <sub>JA</sub>	38°C/W
θ <sub>JC</sub>	4°C/W
Junction Temperature Range	
Lead Temperature (Soldering, 10 sec.)	260°C
Storage Temperature Range	65°C to 150°C

## **Recommended Operating Conditions**

SVIN, NTC	3V to 5.5V
BAT, LX, STAT, ACOKB, BD, ENB, ILIM	0.3V to 16V
BATIN, ITERM	0.3V to 3.6V
LX Pin current continuous	5A
Junction Temperature Range	-40°C to 125°C
Ambient Temperature Range	



## **Electrical Characteristics**

 $T_A \!\!=\!\! 25^\circ C, \, V_{IN} \!\!=\!\! 5V, \, GND \!\!=\!\! 0V, \, C_{IN} \!\!=\!\! 4.7 uF, \, L \!\!=\!\! 0.68 uH, \, R_{ICHG} \!\!=\!\! 0, \! R_{ITERM} \!\!=\!\! 10 k\Omega, \, unless \, otherwise \, specified.$ 

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
<b>Bias Suppl</b>	y (V <sub>SVIN</sub> )		•			
V <sub>SVIN</sub>	Supply voltage		3		16	V
V <sub>UVLO</sub>	V <sub>SVIN</sub> under voltage lockout threshold	$V_{SVIN}$ rising and measured from $V_{SVIN}$ to GND		2.9	V	
$\Delta V_{\rm UVLO}$	V <sub>SVIN</sub> under voltage lockout hysteresis	Measured from V <sub>SVIN</sub> to GND		100		mV
V <sub>OVP</sub>	Input overvoltage protection	$V_{SVIN}$ rising and measured from $V_{SVIN}$ to GND	5.8			V
$\Delta V_{\text{OVP}}$	Input overvoltage protection hysteresis	Measured from $V_{\mbox{\scriptsize SVIN}}$ to GND		0.5		V
Quiescent	Current					
I <sub>BAT</sub>	Battery discharge current	Shutdown IC			25	uA
I <sub>IN</sub>	Input quiescent current	Disable Charge			1.5	mA
Oscillator a		·	•			
f <sub>SW</sub>	Switching frequency			1000		kHz
T <sub>MINOFF</sub>	Main N-FET minimum off time	With 18V rating		100		ns
T <sub>MAXOFF</sub>	Main N-FET maximum off time	With 18V rating		30		us
T <sub>MINON</sub>	Main N-FET minimum on time	With 18V rating		100		ns
Power MO						
R <sub>NFET_M</sub>	R <sub>DS(ON)</sub> of Main N-FET			80		mΩ
R <sub>NFET_R</sub>	R <sub>DS(ON)</sub> of Rectified N-FET			40		mΩ
R <sub>NFET_B</sub>	R <sub>DS(ON)</sub> of Blocking N-FET			40		mΩ
Voltage Re			I			
V <sub>CV</sub>	2-Cell CV charge mode voltage		8.32	8.40	8.48	V
$\Delta V_{RCH}$	2-Cell Recharge Voltage		100	200	300	mV
V <sub>TRK</sub>	2-cell TC charge mode battery voltage threshold	V <sub>BAT</sub> rising edge threshold	5.4	5.6	5.8	V
Battery Co	nnect Detection					
	NTC voltage threshold for Battery					
$V_{DET}$	detect	NTC Falling Edge	85%		95%	V <sub>SVIN</sub>
t <sub>DET</sub>	Detect delay time			260		ms
Charge Cu				200		1115
Charge Cu	Internal charge current accuracy for Constant Current Mode	I <sub>CC</sub> =1000mA	-10%		10%	
	Internal charge current accuracy for Trickle Current Mode	I <sub>TC</sub> =100mA	-50%		50%	
Output Vo		1	1			1
V <sub>OVP</sub>	Output voltage OVP threshold		105%	110%	115%	V <sub>CV</sub>
Input Curi	1 0	1	10,570	11070	11,370	• CV
input Curi		Float ILIM		250		
V <sub>DISS</sub>	Vin drop for slow CC REF	Pull low ILIM		375		mV
▼ DISS	discharge Voltage Threshold	Pull high ILIM		500		ΠV
A 17	Slow Discharge Voltege Hystores			500		mV
$\Delta V_{\text{DISS}}$	Slow Discharge Voltage Hysteresis	Positive edge				mV
N/	Vin drop for fast CC REF discharge	Float ILIM		500		
V <sub>DISF</sub>	Voltage Threshold	Pull low ILIM		750		mV
4 \$ 7	6	Pull high ILIM		1000		
$\Delta V_{\text{DISF}}$	Fast Discharge Voltage Hysteresis			50		mV
Terminatio	on Current					



SILA						
I <sub>TERM</sub>	Termination current threshold	R <sub>ICHG</sub> =0,R <sub>ITERM</sub> =10k R <sub>ICHG</sub> =5k,R <sub>ITERM</sub> =5k		10% 20%		Icc
		$R_{ICHG}=7.5k, R_{ITERM}=2.5k$		40%		icc
Timer		·				
T <sub>MC</sub>	Charge mode change delay time			30		ms
T <sub>TERM</sub>	Termination delay time			30		ms
t <sub>RCH deg</sub>	Recharge deglitch time			15		S
Short Cire	cuit Protection					
V <sub>SHORT</sub>	Output short protection threshold		1.70	2.00	2.30	V
Linear ch	arger Mode					
I <sub>LCHG</sub>	Battery Charger current when the blocking FET is in linear mode	V <sub>BAT</sub> <v<sub>SHORT</v<sub>		5%		I <sub>CC</sub>
I <sub>LPEAK</sub>	Peak linear current when Battery is absent			1		А
V <sub>BD</sub>	Bus voltage regulation		5.8	6	6.2	V
V <sub>TRON</sub>	Blocking FET fully turn on threshold V <sub>TRON</sub> =V <sub>BAT</sub> -V <sub>IN</sub>	$V_{BAT} > V_{TRK}$		100		mV
Enable O	N/OFF Control	•				
V <sub>ENH</sub>	High level logic for enable off		1.5			V
V <sub>ENL</sub>	Low level logic for enable on				0.4	V
	hermal Protection NTC	•				
· · · ·	Under temperature protection		70%	75%	80%	
UTP	Under temperature protection hysteresis	Falling edge		5%		<b>X</b> 7
	Over temperature protection		28%	30%	32%	V <sub>SVIN</sub>
OTP	Over temperature protection hysteresis	Rising edge		2%		
Thermal I	Regulation And Thermal shutdown	•	•			
T <sub>REG</sub>	Thermal regulation threshold			120		°C
T <sub>REGHYS</sub>	Thermal regulation hysteresis falling edge			20		°C
	Thermal regulation fold back ratio			0.25		I <sub>CC</sub>
T <sub>SD</sub>	Thermal shutdown temperature	Rising Threshold		160		°C
T <sub>SDHYS</sub>	Thermal shutdown temperature hysteresis			30		°C

**Note 1**: Stresses beyond the "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note 2:  $\theta_{JA}$  is measured in the natural convection at  $T_A = 25^{\circ}C$  on a low effective four-layer thermal conductivity test board of JEDEC 51-3 thermal measurement standard.

Note 3: The device is not guaranteed to function outside its operating conditions



## **General Function Description**

SY20762F is a 3.0-5.5V<sub>IN</sub>, 2A two-cell synchronous boost Li-Ion battery charger integrates 1MHz switching frequency and full protection functions. The charge current up to 2A can be programmed by using the external resistor for different portable applications and indicates the charger current information simultaneous. It has adaptive input current limit with selectable threshold for safety battery charge operation. SY20762F can disconnect output when there is output short circuit or shutdown happens. It consists of 18V rating FETs with extremely low ON resistance to achieve high charge efficiency and simple peripheral circuit design.

#### **Charging Status Indication Description**

- 1. Charge-In-Process Pull and keep STAT pin to Low;
- 2. Charge Done Pull and keep STAT pin to High;
- **3.** Fault Mode Output high and low voltage alternatively with 1.3Hz frequency. Connect a LED from SVIN to STAT pin, LED ON means Charge-in-Process, LED OFF means Charge Done, LED Flashing with 1.3Hz means Fault Mode.

STAT pin will be open internally if the adapter is absent.

## **Input and Battery Indication Description**

SY20762F has input and battery detecting functions. If the input is present, the ACOKB will be pulled down. Same, the device can judge if the battery pack is inserted based on the NTC pin voltage. If the battery is present, BATIN can be pulled high thru the pull-up resister.

ACOKB and BATIN detection always work if the adapter is present, regardless of the ENB level.

ACOKB and BATIN pins will be open internally if the adapter is absent.

## Switching Mode Boost Charger Basic Operation Description

#### Switching Mode Control Strategy

SY20762F is a switching mode Boost charger for the applications with USB power input. SY20762F utilizes quasi-fixed frequency constant OFF time control to simplify the internal close-loop compensation design. Slope compensation is not necessary for the stable operation. The quasi-fixed frequency settled at 1MHz is easy for the size minimization of peripheral circuit design. During the light load operation, the OFF time is going to be stretched to achieve frequency fold back.

#### **Operation Principle**

SY20762F can normally work with or without Li-Ion battery both.

#### **Battery Present**

Before SY20762F start-up,  $C_{BD}$  is charged by the battery thru the body diode of blocking FET, and  $V_{BD}$  equals to  $V_{BAT}$ .

If the plug in input voltage V<sub>IN</sub> is higher than  $V_{BD}=V_{BAT}$ ,  $C_{BD}$  is charged by  $V_{IN}$  further thru the body diode of sync-FET. Under this condition, the Boost charger operates in light load mode and regulates the V<sub>BD</sub> at 6V and the blocking FET works in linear charge mode. If the V<sub>BAT</sub> is lower than the internal short circuit threshold V<sub>SHORT</sub>, the linear charge current is  $1/20\ I_{CC}.$  When  $V_{BAT}$  is higher than  $V_{SHORT}$  but lower than the threshold of trickle charge, the linear charge current is 1/10 of I<sub>CC</sub>. Note that, charging current would not be increased to I<sub>CC</sub> when the block FET operates in linear mode. With the increasing of  $V_{BAT}$ , when  $V_{BAT}$  is higher than both  $V_{IN}$  and  $V_{TRK}$  the blocking FET is fully turned on and the switching mode boost charger takes over the battery charging. The current in the blocking FET is mirrored to be as the charging current  $I_{CHG}$ . If  $V_{IN}$  is lower than  $V_{BD}=V_{BAT}$  at the plug in time, the switching mode boost charger starts work directly.

During the charging mode, constant (trickle) charging current loop is active first. When  $V_{BAT}$  equals to constant voltage threshold  $V_{CV}$ , constant voltage loop takes over and pull down the charging current. When  $I_{CHG}$  is lower than the termination current threshold  $I_{TERM}$ , the main FET of boost charger is turned off firstly. Sync-FET and blocking FETs are turned off together when the current is down to zero. Then, SY20762F is waiting for recharge mode.

#### **Battery Absent**

If there's no battery connection detected thru NTC pin, SY20762F operates as a normal switching mode boost converter. When  $V_{SVIN}$  is higher than UVLO threshold, the blocking FET is softly turned on. After the blocking FET fully turn-on, switching mode boost converter starts work. The internal current loop and voltage loop are active both.

#### **Basic Protection Principle**

SY20762F has fully battery charging protection. When the input over voltage protection, the output over voltage protection or the thermal protection happens, the main FET of the boost charger is turned off immediately. The sync-FET and the blocking FET are



turned off later when the current is down to zero. When the  $V_{BAT}$  is lower than  $V_{SHORT}$ , the short circuit protection happens. The main FET is turned off firstly. The block FET enters linear mode with 1/20 I<sub>CC</sub> charging current. When  $V_{BAT}$  recovers back to be higher than  $V_{SHORT}$ , the boost charger restarts to work at light load and regulates  $V_{BD}$  at 6V. The linear charge current is increased from 1/20 I<sub>CC</sub> to trickle current.

#### **Basic Adaptive Input Current Limit Principle**

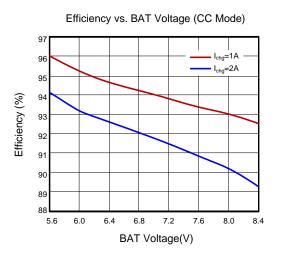
SY20762F has adaptive input current limit function. Before the IC starts charging work, the input voltage is detected and saved as reference  $V_{INREF}$ . Once IC starts to charge, the output charging current  $I_{CHG}$  is ramped up softly and the  $V_{SVIN}$  drop is monitored

simultaneously. When the input voltage drop is larger than V<sub>DISS</sub> the output charging current reference I<sub>CHGREF</sub> starts to be discharged slowly and when the voltage drop is larger than V<sub>DISF</sub> the I<sub>CHGREF</sub> starts to be fast discharged. With the discharging of I<sub>CHGREF</sub>, the charging current is deceased and the V<sub>SVIN</sub> would recover back. Once the V<sub>SVIN</sub> recovers back into the normal range, the I<sub>CHGREF</sub> is kept on the current value. The I<sub>CHGREF</sub> would be decreased along with the increasing of output voltage to keep the input power at the maximum value. The internal digital machine state is built up to achieve this function.

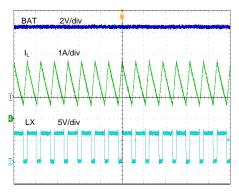


# **Typical Performance Characteristics**

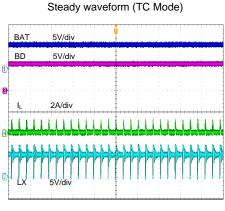
 $T_A=25^{\circ}C, V_{IN}=5V, L=0.68uH, R_{ICHG}=0, R_{ITERM}=10k\Omega, unless otherwise specified.$ 



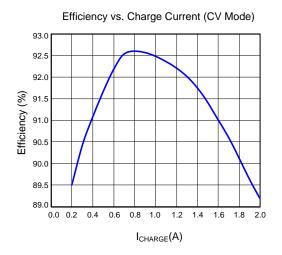
Steady waveform (CC Mode)



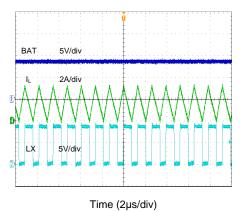
Time (2µs/div)





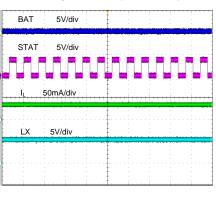


Steady waveform (CV Mode)



nine (zµs/div)

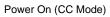
Steady waveform (Short Mode)

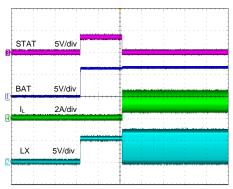


Time (1s/div)

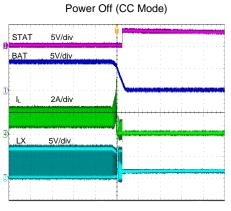


# SY20762F

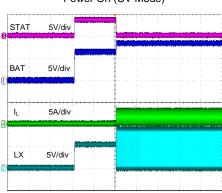




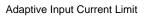


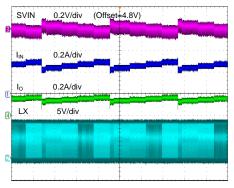


Time (2ms/div)

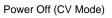


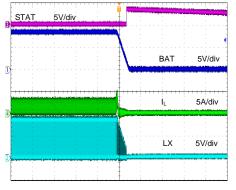
Time (400ms/div)



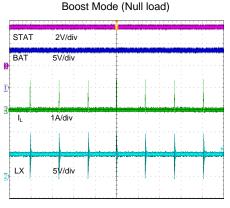








Time (4ms/div)



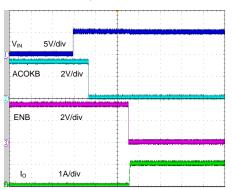
Time (20µs/div)

# Time (400ms/div) Power On (CV Mode)

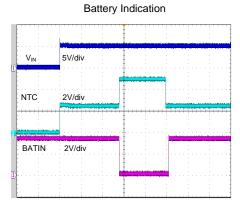


# SY20762F

Input Indication



Time (1s/div)



Time (2s/div)



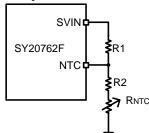
## **Applications Information**

Because of the high integration of SY20762F, the application circuit based on this regulator IC is rather simple. Only input capacitor  $C_{IN}$ , output capacitor  $C_{OUT}$ , inductor L, NTC resistors R1, R2 need to be selected for the targeted applications specifications.

#### NTC resistor:

SY20762F monitors battery temperature by measuring the input voltage and NTC voltage. The controller triggers the UTP or OTP when the rate K (K=  $V_{\text{NTC}}/V_{\text{SVIN}}$ ) reaches the threshold of UTP (K<sub>UT</sub>) or OTP (K<sub>OT</sub>). The temperature sensing network is showed as below.

Choose R1 and R2 to program the proper UTP and OTP points.



The calculation steps are:

- 1. Define KUT,  $K_{UT} = 70 \sim 80\%$
- 2. Define Kot, Kot = 28~32%
- 3. Assume the resistance of the battery NTC thermistor is Rut at UTP threshold and Rot at OTP threshold.

4. Calculate R2,

$$R2 = \frac{K_{OT}(1 - K_{UT})R_{UT} - K_{UT}(1 - K_{OT})R_{OT}}{K_{UT} - K_{OT}}$$

5. Calculate R1

$$R1 = (1 / Kot - 1)(R2 + Rot)$$

If choose the typical values  $K_{\rm UT}$  =75% and Kot=30%, then

$$R2 = 0.17 R_{UT} - 1.17 R_{OT}$$
  
R1 = 2.3(R2 + R\_{OT})

#### Input capacitor CIN:

The ripple current through input capacitor is greater than

$$I_{\text{CIN\_RMS}} = \frac{V_{\text{IN}} * (V_{\text{OUT}} - V_{\text{IN}})}{2\sqrt{3} * L^* F_{\text{SW}} * V_{\text{OUT}}}$$

X5R or X7R ceramic capacitors with greater than 4.7uF capacitance are recommended to handle this ripple current.

## **Output capacitor C**OUT:

The output capacitor is selected to handle the output ripple noise requirements. This ripple voltage is related to the capacitance and its equivalent series resistance (ESR). For the best performance, it is recommended to use X5R or better grade low ESR ceramic capacitor. The voltage rating of the output capacitor should be higher than the maximum output voltage. The minimum required capacitance can be calculated as:

$$C_{OUT} = \frac{I_{CC*}(V_{OUT} - V_{IN})}{F_{SW} \times V_{OUT} \times V_{RIPPLE}}$$

 $V_{\text{RIPPLE}}$  is the peak to peak output ripple,  $I_{\text{CC}}$  is the setting charge current.

For SY20762F, output capacitor is paralleled by  $C_{BD}$  and  $C_{BAT}$ , for smaller output ripple noise, each capacitor with greater than 10uF capacitance is recommended.

#### Inductor L:

There are several considerations in choosing this inductor.

 Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the average input current. The inductance is calculated as:

$$L = \left(\frac{V_{\text{IN}}}{V_{\text{OUT}}}\right)^2 \frac{(V_{\text{OUT}} - V_{\text{IN}})}{I_{\text{CC}} \times F_{\text{SW}} \times 40\%}$$

Where  $F_{SW}$  is the switching frequency and  $I_{CC}$  is the setting charge current.

The SY20762F is quite tolerant of different ripple current amplitude. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

2) The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

$$I_{SAT,MIN} > \left(\frac{V_{OUT}}{V_{IN}}\right) \times I_{CC} + \left(\frac{V_{IN}}{V_{OUT}}\right) \times \frac{(V_{OUT} - V_{IN})}{2 \times F_{SW} \times L}$$

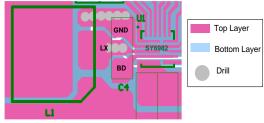
3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with DCR<10mohm to achieve a good overall efficiency.



#### Layout Design:

The layout design of SY20762F regulator is relatively simple. For the best efficiency and minimum noise problems, we should place the following components close to the IC: C<sub>SVIN</sub>, L, C<sub>BD</sub>. 1) The loop of main MOSFET, rectifier diode, and

 The loop of main MOSFET, rectifier diode, and C<sub>BD</sub> must be as short as possible



2) It is desirable to maximize the PCB copper area connecting to GND pin to achieve the best thermal and noise performance.

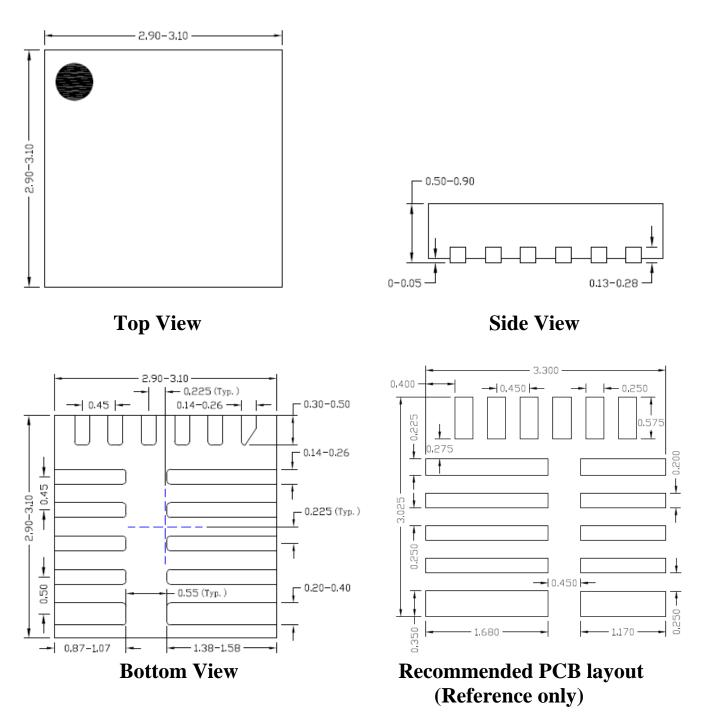
3)  $C_{SVIN}$  must be close to pin SVIN and GND.

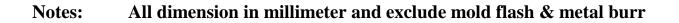
4) The PCB copper area associated with LX pin must be minimized to avoid the potential noise problem.

5) The small signal component R<sub>ICHG</sub> must be placed close to IC and must not be adjacent to the LX net on the PCB layout to avoid the noise problem





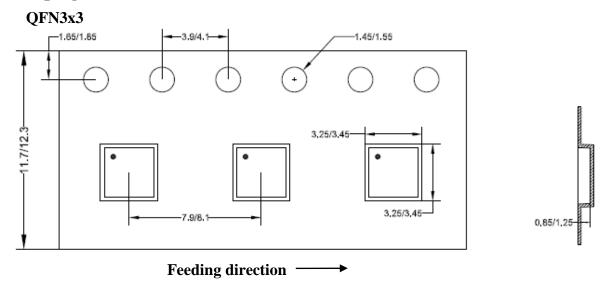




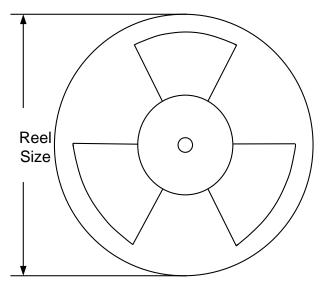




## 1. Taping orientation



## 2. Carrier Tape & Reel specification for packages



Package	Tape width	Pocket	Reel size	Trailer	Leader length	Qty per
types	(mm)	pitch(mm)	(Inch)	length(mm)	(mm)	reel
QFN3x3	12	8	13"	400	400	5000

## 3. Others: NA



# SY20762F

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