

# High Efficiency, 4uA Ultra Low $I_Q$ , 3MHz, 0.6A Inductor Built-in Synchronous Step Down Regulator

# **General Description**

The SY20681 is a 3MHz, 0.6A,  $4\mu A$  ultra low quiescent current, synchronous step-down converter which integrates an inductor and a control IC in one tiny package (2.0mm×1.5mm, H=1.0mm). It can operate over a wide input voltage range from 1.8V to 5.5V and integrates main switch and synchronous switch with very low  $R_{DS\,(ON)}$  to minimize the conduction loss.

### **Features**

- Low  $R_{DS(ON)}$  for Internal Switches (Top/Bottom) :  $230m\Omega/110m\Omega$
- Integrate an Inductor to Minimizes the External Components and PCB Layout Design
- 1.8~5.5V Input Voltage Range
- 4μA Ultra Low Quiescent Current
- High Switching Frequency 3MHz Minimizes the External Components
- Internal Soft-start Limits the Inrush Current
- 100% Dropout Operation
- RoHS Compliant and Halogen Free
- Compact Package: QFN2x1.5-8

### **Applications**

- Mobile Phone, Smart Phone
- Bluetooth Headsets
- Portable Game Console
- Digital Camera, Camcorder

# **Typical Applications**

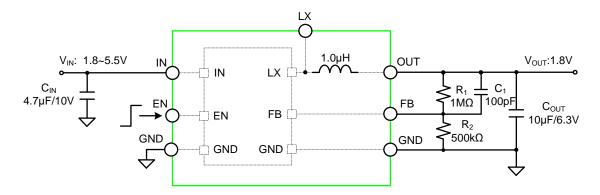


Figure 1. Schematic Diagram

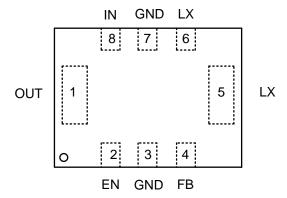


# **Ordering Information**

Ordering Part Number	Package type	Top Mark
SY20681QUC	QFN2×1.5-8 RoHS Compliant and Halogen Free	Kaxyz

x=year code, y=week code, z= lot number code

# Pinout (top view)



Pin Name	Pin Number	Pin Description		
OUT	1	Output Pin. Decouple this pin to ground with at least a 10uF ceramic capacitor.		
EN	2	Enable control. Pull high to turn on. Do not leave it floating.		
GND	3, 7	Ground pin.		
FB	4	Output adjustable version. Connect this pin to the center point of the output resistor divider to program the output voltage: $V_{OUT}=0.6\times(1+R_1/R_2)$ .		
LX	5,6	Built-in inductor node. Leave it floating.		
IN	8	Input pin. Decouple this pin to GND pin with at least a 4.7uF ceramic capacitor.		





Absolute Maximum Ratings (1)	Min	Max	Unit
IN,LX,OUT		6	V
FB	-0.3	IN + 0.3	
Junction Temperature, Operating	-40	150	
Lead Temperature (Soldering, 10sec.)		260	°C
Storage Temperature	-65	150	

Thermal Information (2)		Max	Unit
θ <sub>JA</sub> Junction-to-ambient Thermal Resistance		60.3	°C/W
θ <sub>JC</sub> Junction-to-case Thermal Resistance			
P <sub>D</sub> Power Dissipation T <sub>A</sub> =25°C		400	mW

Recommended Operating Conditions (3)  IN		Max	Unit
IN	1.8	5.5	V
Junction Temperature	-40	125	°C
Ambient Temperature	-40	85	°C



## **Electrical Characteristics**

<b>Electrical Characteristics</b> $V_{IN} = 4.2V$ , $V_{OUT} = 1.8V$ , $C_{OUT} = 10\mu F$ , $T_A = 25^{\circ}C$ , unless otherwise specified							
Parameter	Symbol	<b>Test Conditions</b>	Min	Тур	Max	Unit	
Input Voltage Range	$V_{\rm IN}$		1.8		5.5	V	
Quiescent Current	$I_Q$	$I_{OUT}=0, V_{FB}=V_{REF}\times 105\%$		4		μΑ	
Shutdown Current	$I_{SHDN}$	EN=0		0.1	1	μΑ	
Feedback Reference Voltage	$V_{REF}$		0.588	0.6	0.612	V	
PFET R <sub>ON</sub>	R <sub>DS(ON)</sub> ,P			230		mΩ	
NFET RON	R <sub>DS(ON)</sub> ,N			110		mΩ	
Inductance	L			1.0		μΗ	
PFET Current Limit	$I_{LIM}$		1.3			A	
EN Rising Threshold	$V_{\mathrm{ENH}}$		1.2			V	
EN Falling Threshold	$V_{\mathrm{ENL}}$				0.4	V	
Input UVLO Threshold	$V_{\rm UVLO}$				1.8	V	
UVLO Hysteresis	V <sub>HYS</sub>			0.1		V	
Oscillator Frequency	fosc			3		MHz	
Min ON Time				80		ns	
Max Duty Cycle			100			%	
Soft-start Time	$t_{SS}$			1		ms	
Thermal Shutdown Temperature	$T_{\mathrm{SD}}$			150		°C	
Thermal Shutdown Hysteresis	T <sub>HYS</sub>			15		°C	
Output discharge resistor	R <sub>DSC</sub>			70		Ω	

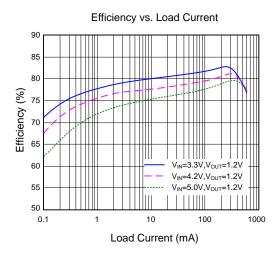
**Note 1**: Stresses beyond the "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

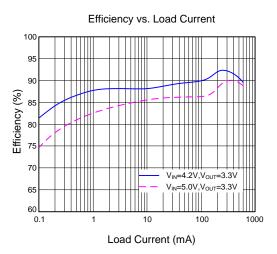
**Note 2**:  $\theta_{JA}$  is measured in the natural convection on a two-layer Silergy Evaluation Board.

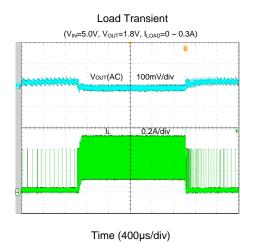
**Note 3:** The device is not guaranteed to function outside its operating conditions

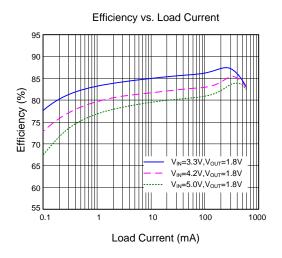


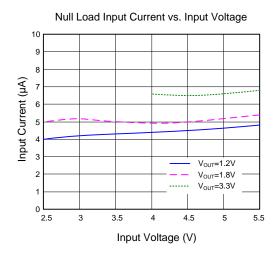
# **Typical Performance Characteristics**

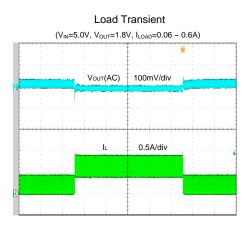






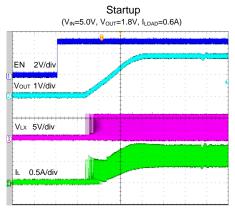






Time (400µs/div)

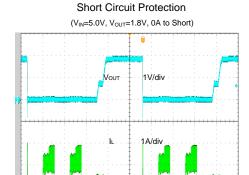




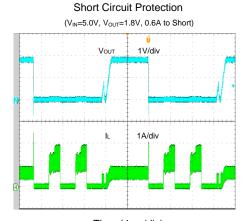


# (V<sub>IN</sub>=5.0V, V<sub>OUT</sub>=1.8V, I<sub>LOAD</sub>=0.6A) EN 2V/div Vout 1V/div V<sub>LX</sub> 5V/div Time (100µs/div)

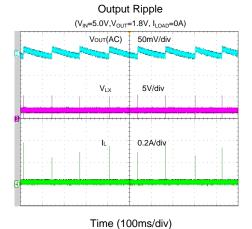
Shutdown

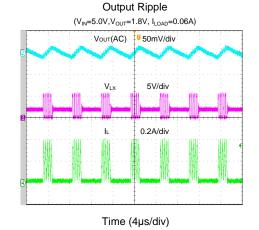


Time (4ms/div)



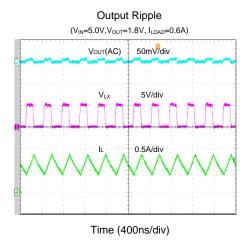
Time (4ms/div)

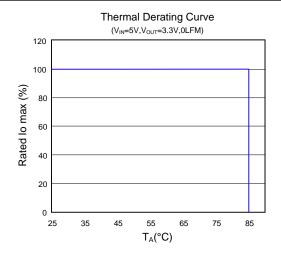




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### **Notes:**

- 1) T<sub>A</sub>: Air temperature, 0.5 inch above the IC.
- 2) Based on a four-layer Silergy Evaluation Board in the natural convection.
- 3) The inductor temperature is not beyond  $115^{\circ}$ C under this TD curve.
- 4) For customer's specific application, the recommended inductor temperature limitation is 115°C.



## **Operation**

SY20681 is a 3MHz, 0.6A, 4µA ultra low quiescent current, synchronous step-down converter which integrates an inductor and a control IC in one tiny package (2.0mm×1.5mm, H=1.0mm). It can operate over a wide input voltage range from 1.8V to 5.5V and integrates main switch and synchronous switch with very low R<sub>DS (ON)</sub> to minimize the conduction loss.

### **Applications Information**

Because of the high integration in SY20681, the application circuit based on this IC is rather simple. Only the input capacitor C<sub>IN</sub>, the output capacitor C<sub>OUT</sub> and the feedback resistors (R<sub>1</sub> and R<sub>2</sub>) need to be selected for the targeted application specifications.

### Feedback Resistor Dividers R1 and R2:

Choose R<sub>1</sub> and R<sub>2</sub> to program the proper output voltage. To minimize the power consumption under light loads, it is desirable to choose large resistance values for both R<sub>1</sub> and R<sub>2</sub>. A value of greater than  $1M\Omega$  is highly recommended for  $R_1$ resistor. If  $R_1 = 1M\Omega$  is chosen, then  $R_2$  can be calculated to be:

$$R_2 = \frac{0.6V}{(V_{OUT} - 0.6V)} \times R_1$$

### **Input Capacitor CIN:**

# **Typical Application Schematic**

LX 1.0uH  $V_{\text{OUT}}$ ΙN OUT IN LX FΒ  $C_{OUT}$ ΕN FΒ **GND** GND **GND** 

# **Recommended Table**

V <sub>OUT</sub> (V)	$R_1(k\Omega)$	$\mathbf{R}_2(\mathbf{k}\Omega)$
1.2	1000	1000
1.8	1000	500
3.3	1000	221

A typical X7R or better grade ceramic capacitor greater than 4.7µF capacitance is recommended. To minimize the potential noise problem, this ceramic capacitor should be placed really close to the IN and GND pins. Care should be taken to minimize the loop area formed by C<sub>IN</sub>, and IN/GND pins.

### **Output Capacitor Cout:**

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For the best performance, it is recommended to use an X7R or better grade ceramic capacitor with 6V rating and greater than 10uF capacitance, when V<sub>OUT</sub> is 1.2V, C<sub>OUT</sub> is recommended to be greater than 20μF.

### **Load Transient Considerations:**

SY20681 integrates the compensation components to achieve good stability and fast transient response. In some applications, adding a 22pF~100pF ceramic cap in parallel with R1 may further speed up the load transient responses and is thus recommended for applications with large load transient step requirements.

### **Short Circuit Protection:**

SY20681 integrates hic-cup mode hard short protection function. If the FB is below 0.2V, the internal soft-start node and the error amplifier output will be reset immediately. The IC works in hic-cup protection mode. The hiccup frequency is about 200Hz, and the hic-cup duty is 50%. If the hard short condition is removed, the IC will go back to normal operation.



### **Bom List**

Reference Designator	Description	Part Number	Manufacturer
$C_{IN}$	4.7µF/10V, 0402, X5R	C1005X5R1A475M	TDK
$C_{OUT}$	10μF/6.3V, 0603, X5R	C1608X5R0J106M	TDK
$C_1$	100pF/50V, 0603, C0G	C1608C0G1H101J	TDK
$R_1$	$1M\Omega$ , 1%, 0603		
$R_2$	500kΩ, 1%, 0603		
$R_3$	$1M\Omega, 0603$		

### **Layout Design:**

For the minimum noise problems, we should place the following components close to the IC:  $C_{IN}$  and  $C_{OUT}$ .

- 1) It is desirable to maximize the PCB copper area connecting to the GND pin to achieve the best thermal and noise performance. If the board space allowed, a ground plane is highly desirable.
- 2) The  $C_{\rm IN}$  must be close to the IN and GND pins. The loop area formed by the  $C_{\rm IN}$  and GND must be minimized.
- 3) Connect the LX pins together to reduce the inductor DCR. It is strongly recommended to reduce the LX routing area to avoid the potential noise problem.
- 4) The trace connecting to the FB pin must NOT be adjacent to the LX node on the PCB layout to minimize the noise coupling to the FB pin.

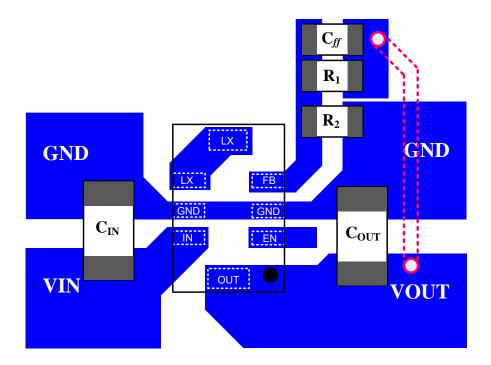
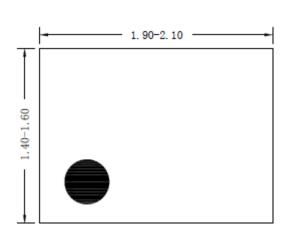
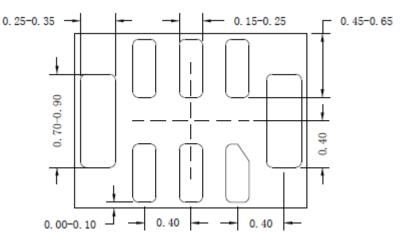


Figure 2. PCB Layout Suggestion



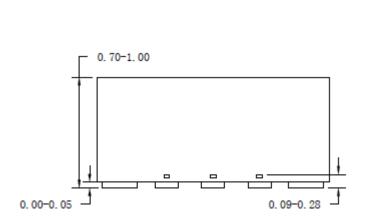
# QFN2×1.5-8 Package Outline Drawing

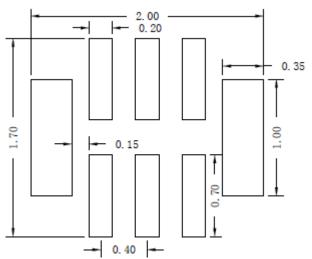




**Top View** 

**Bottom View** 





**Side View** 

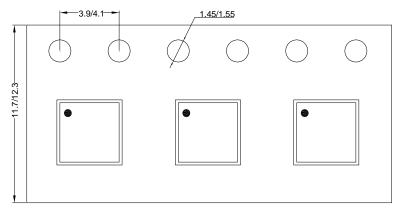
Recommended PCB layout (Reference only)

Notes: All dimension in millimeter and exclude mold flash & metal burr



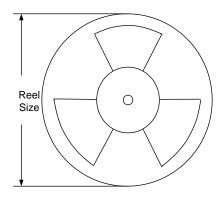
# **Taping & Reel Specification**

# 1. QFN2×1.5 taping orientation



Feeding direction ----

# 2. Carrier Tape & Reel specification for packages



Package type	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Trailer length(mm)	Leader length (mm)	Qty per reel
QFN2×1.5	8	4	7	400	160	3000

## 3. Others: NA



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