

### General Description

The SY24634 includes a high common-mode, current-sensing amplifier, as well as a high-speed comparator that can be configured to detect under-current conditions. The Current Sense Amplifier is used to accurately measure a small voltage developed across a shunt resistor and provide a proportional analog value at its output.

The built-in comparator measures the voltage developed at the output of the Current Sense Amplifier and compares that voltage with an externally configured threshold. When the voltage at the Amplifier output is lower than the threshold, its open-drain output will be driven low, effectively enabling fast under-current detection in a system. A single external resistor can be used to configure the under-current threshold. This current-shunt monitor can measure differential voltage signals on common-mode voltages that can vary from 0V up to 36V, independent of the supply voltage.

The open-drain Alert Output can be configured to operate in either transparent mode or latched mode. The device alert response time is under 3μs, which allows for quick detection of under-current events.

The SY24634 operates from a single 2.7V to 5.5V supply, drawing a maximum supply current of 400μA. The device is specified over the extended operating temperature range (-40°C to +125°C), and is available in an 8-pin MSOP package.

### Features

- Wide Common-mode Range: 0V to 36V
- Dual Output: Amplifier and Comparator Output
- High-Accuracy Amplifier:
  - Offset Voltage: 35μV (Max)
  - Offset Voltage Drift: 0.5μV/°C (Max)
  - Gain Error: 0.2% (Max)
  - Gain Error Drift: 10ppm/°C (Max)
- Available Amplifier Gain: 100V/V
- Programmable Alert Threshold Setting through a Single Resistor
- Total Alert Response Time: 3μs
- Open-drain Output with Latching Mode
- Package: MSOP8

### Applications

- Under-current Protection
- Power-Supply Protection
- Circuit Breakers
- Computers and Servers
- Telecom Equipment
- Battery Management

### Typical Application

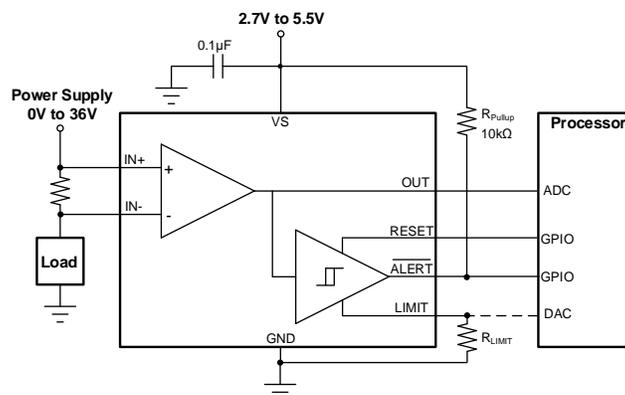


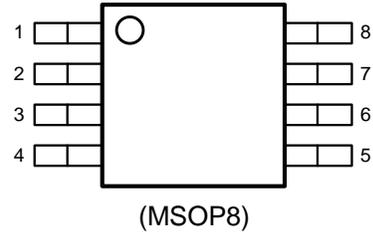
Figure 1. Schematic Diagram

## Ordering Information

Ordering Part Number	Package Type	Top Mark
SY24634CAP	MSOP8	FBYxyz

x = year code, y = week code, z = lot number code

## Pinout (Top View)



## Pin Description

Pin No	Pin Name	Pin Description
1	V <sub>s</sub>	Power supply, 2.7V to 5.5V.
2	OUT	Output voltage.
3	LIMIT	Alert threshold limit input.
4	GND	Ground.
5	RESET	Transparent or latch mode selection input.
6	ALERT	Over limit alert; active-low, open-drain output.
7	IN-	Connect to load side of shunt resistor.
8	IN+	Connect to supply side of shunt resistor.

## Absolute Maximum Ratings

Parameter (Note 1)	Min	Max	Unit
V <sub>s</sub>	-0.3	6	V
V <sub>IN+</sub> - V <sub>IN-</sub> (Differential)	-40	40	
V <sub>IN+</sub> , V <sub>IN-</sub> (Common-Mode)	-0.3	40	
Alert Output V <sub>ALERT</sub>	-0.3	6	
Input Voltage at Any Pin	-0.3	V <sub>s</sub> +0.3	
Maximum Junction Temperature		150	°C
Storage Temperature Range	-65	150	
ESD: HBM (Human Body Model)	± 2000		V
ESD: CDM (Charged Device Model)	± 1000		V

## Thermal Information

Parameter (Note 2)	Max	Unit
θ <sub>JA</sub> Junction-to-Ambient Thermal Resistance	175	°C/W
θ <sub>JC</sub> Junction-to-Case Thermal Resistance (top)	40	

## Recommended Operating Conditions

Parameter (Note 3)	Min	Max	Unit
V <sub>s</sub>		5	V
Common-mode, V <sub>IN+</sub> , V <sub>IN-</sub>		12	
Operation Temperature	-40	125	°C

## Electrical Characteristics

$T_A = 25^\circ\text{C}$ ,  $V_{IN} = V_{IN+} - V_{IN-} = 10\text{mV}$ ,  $V_S = 5\text{V}$ ,  $V_{IN+} = 12\text{V}$  and  $V_{LIMIT} < 1\text{V}$ , unless otherwise noted.

Parameter (Note 4)	Symbol	Conditions	Min	Typ	Max	Units
<b>Input</b>						
Input Voltage	$V_{CM}$		0		36	V
Differential Input Voltage	$V_{IN}$	$V_{IN} = V_{IN+} - V_{IN-}$	0		50	mV
Common-Mode Rejection	CMR	$V_{IN} = 0\text{V to } 36\text{V}$ , $T_A = -40^\circ\text{C to } +125^\circ\text{C}$	110	120		dB
Offset Voltage, RTI (Note 5)	$V_{OS}$			$\pm 10$	$\pm 35$	$\mu\text{V}$
Offset Voltage Drift, RTI	$dV_{OS}/dT$	$T_A = -40^\circ\text{C to } +125^\circ\text{C}$		0.1	0.5	$\mu\text{V}/^\circ\text{C}$
Power Supply Rejection Ratio	PSR	$V_S = 2.7\text{V to } 5.5\text{V}$ , $V_{IN+} = 12\text{V}$ , $T_A = -40^\circ\text{C to } +125^\circ\text{C}$		$\pm 0.1$	$\pm 10$	$\mu\text{V/V}$
Input Bias Current (Note 6)	$I_B$			125		$\mu\text{A}$
Input Offset Current (Note 7)	$I_{OS}$			$\pm 0.1$		$\mu\text{A}$
<b>Output</b>						
Gain	G			100		V/V
Gain Error		$V_{OUT} = 0.5\text{V to } V_S - 0.5\text{V}$		$\pm 0.11$	$\pm 0.2$	%
		$T_A = -40^\circ\text{C to } +125^\circ\text{C}$		3	10	ppm/ $^\circ\text{C}$
Nonlinearity Error		$V_{OUT} = 0.5\text{V to } V_S - 0.5\text{V}$		$\pm 0.01$		%
Maximum Capacitive Load		No sustained oscillation		1		nF
<b>Voltage Output</b>						
Swing to $V_S$ Power Supply Rail		$R_L = 10\text{k}\Omega$ to GND, $T_A = -40^\circ\text{C to } +125^\circ\text{C}$		$V_S - 0.05$	$V_S - 0.1$	V
Swing to GND		$R_L = 10\text{k}\Omega$ to GND, $T_A = -40^\circ\text{C to } +125^\circ\text{C}$		$V_{GND} + 20$	$V_{GND} + 30$	mV
<b>Frequency Response</b>						
Bandwidth	BW			500		kHz
Slew Rate	SR			4		V/ $\mu\text{s}$
<b>Noise, RTI</b>						
Voltage Noise Density				30		nV/ $\sqrt{\text{Hz}}$
<b>Comparator</b>						
Total Alert Propagation Delay	$t_p$	$V_{LIMIT} = 50\text{mV}$ , Input overdrive = 0.1mV (Note 5) $V_{IN}$ to Alert Propagation		3		$\mu\text{s}$
Limit Threshold Output Current	$I_{LIMIT}$		79.4	80	80.6	$\mu\text{A}$
		$T_A = -40^\circ\text{C to } +125^\circ\text{C}$	78.8		81.2	
Comparator Offset Voltage	$V_{OS}$			1.5	4.5	mV
Hysteresis	HYS			50		mV
High-Level Input Voltage	$V_{IH}$		1.4		$V_S + 0.3$	V
Low-Level Input Voltage	$V_{IL}$		0		0.4	V
Alert Low-Level Output Voltage	$V_{OL}$	$I_{OL} = 3\text{mA}$		70	300	mV
ALERT Terminal Leakage Input Current		$V_{OH} = 3.3\text{V}$		0.1	1	$\mu\text{A}$
Digital Leakage Input Current		$0 < V_{IN} < V_S$		1		$\mu\text{A}$
<b>Power Supply</b>						
Operating Supply Range	$V_S$	$T_A = -40^\circ\text{C to } +125^\circ\text{C}$	2.7		5.5	V
Quiescent Current	$I_Q$	$V_{IN} = 0\text{mV}$ , $T_A = 25^\circ\text{C}$		300	380	$\mu\text{A}$
		$T_A = -40^\circ\text{C to } +125^\circ\text{C}$			400	

**Note 1:** Stresses beyond the “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**Note 2:** Package thermal resistance is measured in the natural convection at  $T_A = 25^\circ\text{C}$  on an 8.5cm x 8.5cm single-layer Silergy Evaluation Board.

**Note 3:** The device is not guaranteed to function outside its operating conditions.

**Note 4:** Production testing is performed at  $25^\circ\text{C}$ ; limits at  $-40^\circ\text{C}$  to  $+125^\circ\text{C}$  are guaranteed by design, test, or statistical correlation.

**Note 5:** RTI = Referred to Input.

**Note 6:** Input bias current is decided by the average of the input currents of the pin IN+ and IN-.

**Note 7:** Input offset current is decided by the error between the input currents of the pin IN+ and IN-.

**Note 8:** Overdrive =  $(V_{\text{LIMIT}} - V_{\text{OUT}})/\text{Gain}$ .

## Application Information

### Operation

The SY24634 is a fixed-gain, zero-drift current-sense amplifier that amplifies the differential voltage across an external shunt resistor to create an output voltage.

The device features a 0V to 36V common-mode independent of its supply voltage in applications that use low-side or high-side configurations. The SY24634 can withstand, without sustaining damage, the full 36V common-mode voltage at the input pins when the supply voltage is removed.

The zero-drift topology enables high-precision measurements with typical input offset voltages as low as 35 $\mu$ V, and with a temperature contribution of only 0.5 $\mu$ V/ $^{\circ}$ C over the full temperature range of -40 $^{\circ}$ C to +125 $^{\circ}$ C. The low total offset voltage of the SY24634 allows smaller current-sensing shunt resistor values to be used, minimizing the power loss across it.

The built-in comparator uses a single external resistor to set the corresponding current threshold level that is used for signaling under-current conditions. Combining the precise measurement of the current-sensing amplifier with the comparator enables creating a fast acting under-current-detection device.

### Alert Output

The ALERT pin is an active-low, open-drain output that is designed to be pulled low when the monitored current is below the configured threshold. This open-drain output pin should be connected to supply voltage through a 10k $\Omega$  pull-up resistor. The ALERT pin can be pulled up to a voltage beyond the supply voltage  $V_S$ , but the value must not exceed 5.5 V.

When the output voltage of the amplifier is higher than the voltage developed at the LIMIT pin, the comparator output will be in the default high state. When the amplifier output voltage is less than the threshold voltage set using the LIMIT pin, the comparator output will become active and be pulled low, indicating that an under-current condition has occurred.

### Alert Mode

Using the RESET pin, the output can be configured to operate in either Transparent or Latched mode. This will change the behavior of the output when the under-current condition is removed, as described in the following paragraphs.

### Transparent Output Mode

The device will be set to Transparent Mode when the RESET pin is pulled low. In this mode, the Alert output changes state following the input signal with respect to the programmed alert threshold. For example, when the differential input signal drops below the alert threshold, the ALERT output pin will be pulled low. As soon as the differential input signal rises above the alert threshold, the output will return to the default high output state.

When connect the ALERT pin to a hardware interrupt input on a microcontroller, the output state change of the ALERT pin in Transparent Mode can be directly detected and under-current conditions can be handled.

### Latch Output Mode

The latch mode is designed to accommodate applications which do not have the functionality available to continuously monitor the state of the output ALERT pin.

The device can be configured to operate in Latch Mode by connecting the RESET pin to a HIGH level. In this mode when an under-current condition is removed, the ALERT pin will not return to the default high state.

To clear the alert, the RESET pin must be pulled low for at least 100ns. Pulling the RESET pin low allows the ALERT pin to return to the default high level, provided that the differential input signal has risen above the alert threshold. If the input signal is still below the threshold limit when the RESET pin is pulled low, the ALERT pin will remain low.

After processing the under-current condition the external controller can return the device to Latch Mode by setting the RESET pin by to logic high.

### Transparent Output vs. Latch Output

The SY24634 is placed into the corresponding output mode by driving the voltage level at the RESET pin, as shown in Table 1.

Table 1. Output Mode Settings

Output Mode	RESET Terminal Setting
Transparent mode	RESET = low
Latch mode	RESET = high

The Alert pin behavior is shown in Figure 2 for the Latch and Transparent Modes.

When  $V_{OUT}$  rises above the  $V_{LIMIT} + \text{Hysteresis}$  threshold for the first time, the RESET pin will be pulled high by the external resistor.

When the RESET pin is pulled high, the device will be set to Latch mode. In this mode the Alert output state will not return to high when the  $V_{OUT}$  rises above the  $V_{LIMIT} + \text{Hysteresis}$  threshold.

When the RESET pin is pulled low, the ALERT pin will return to the default high level, indicating that the  $V_{OUT}$  is above the  $V_{LIMIT} + \text{Hysteresis}$  threshold. When the  $V_{OUT}$  rises above the  $V_{LIMIT} + \text{Hysteresis}$  threshold for the second time, the RESET pin will already be pulled low. The device is set to transparent mode at this point, and the ALERT pin is pulled back high as soon as the  $V_{OUT}$  rises above the  $V_{LIMIT} + \text{Hysteresis}$  threshold.

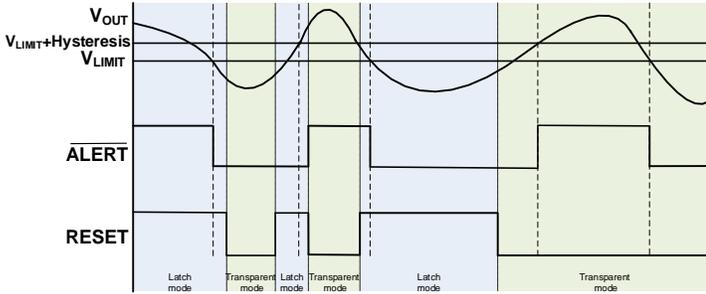


Figure 2. Transparent vs. Latch Mode

### Current-Limit Threshold Setting

The SY24634 will determine if an under-current event is present by comparing the amplified measured voltage developed across the current-sensing shunt resistor with the corresponding voltage configured using the LIMIT pin. The threshold voltage for the LIMIT pin can be set using a single external resistor, or by connecting an external voltage source to the LIMIT pin.

### Resistor-Controlled Current Limit

One approach for setting the limit threshold voltage is to connect a resistor from the LIMIT pin to ground. The value of this resistor  $R_{LIMIT}$  will be chosen to create a corresponding voltage at the LIMIT pin that is equivalent to the output voltage  $V_{OUT}$ , when the maximum desired load current is flowing through the current-sensing resistor. An internal  $80\mu A$  current source is connected to the LIMIT pin to configure the desired threshold by choosing the value of the  $R_{LIMIT}$  resistor.

### Voltage Source Controlled Current Limit

An alternate method for setting the under-current threshold is to connect the LIMIT pin to a programmable DAC (digital-to-analog converter). The benefit of this method is the ability to adjust the threshold, as the system operating conditions change.

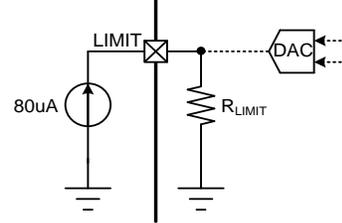


Figure 3. Typical Circuit for Setting  $V_{LIMIT}$

### Hysteresis

The onboard comparator in the SY24634 is designed to reduce the possibility of oscillations when the measured current is near the under-current threshold. A hysteresis level (HYS) of  $50mV$  is utilized for proper operation, as shown in Figure 4.

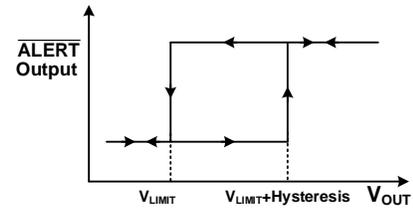
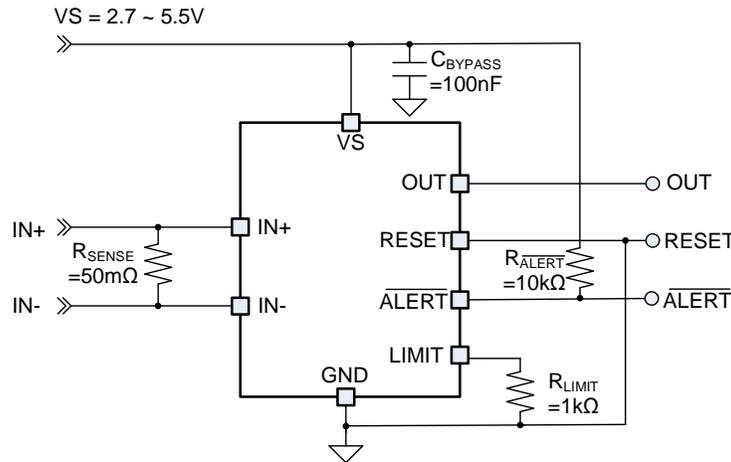


Figure 4.  $V_{LIMIT}$  Hysteresis

## Application Schematic



## BOM List

Designator	Description	Part Number	MFR
C <sub>BYPASS</sub>	100nF/50V/X7R, 0603	GCJ188R71H104KA12D	muRata
R <sub>SENSE</sub>	50mΩ/1W, 1%, 2512	RL2512FK-070R01L	YAGEO
R <sub>LIMIT</sub>	1kΩ, 1%, 0603		
R <sub>ALERT</sub>	10kΩ, 1%, 0603		

## Layout Design

For optimal design, follow these PCB layout considerations:

- Using a Kelvin connection to connect the input pins to the current-sensing resistor  $R_{SENSE}$ . Due to the low resistance values of  $R_{SENSE}$ , poor PCB routing often leads to additional parasitic resistance between input pins, resulting in additional errors that cannot be ignored, this connection technique ensures that only  $R_{SENSE}$  impedance is detected between the input pins. Minimizing the loop formed by these connections.
- Place the bypass capacitor (a 0.1μF MLCC is recommended) very near VCC and GND.
- To limit the generation of additional capacitance on this node, connect the  $R_{LIMIT}$  to the GND pin as directly as possible.

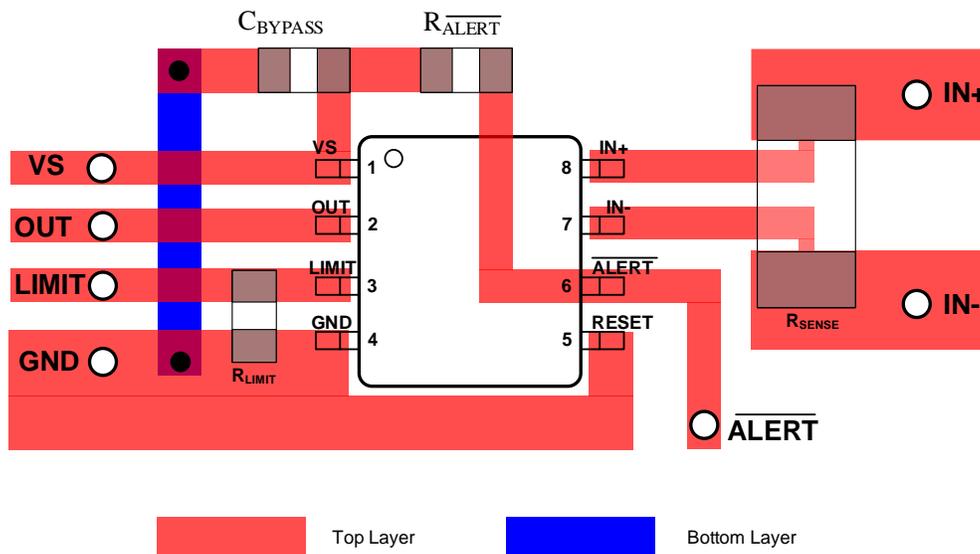
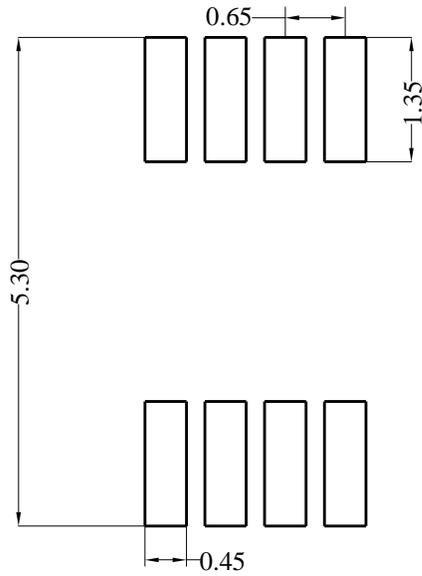
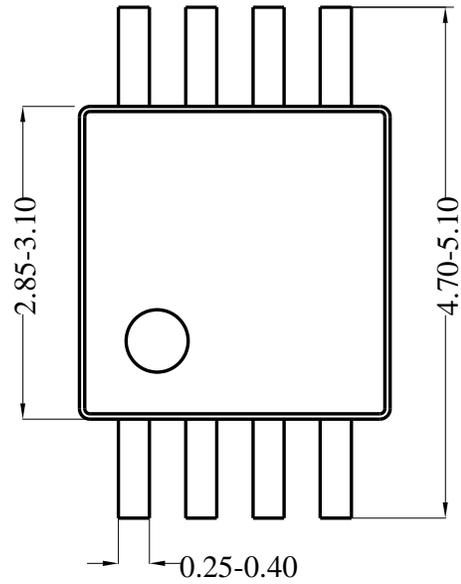


Figure 5. Recommended Layout

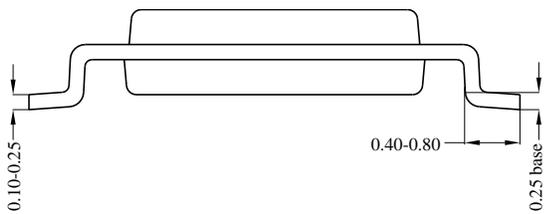
**MSOP8 Package Outline Drawing**



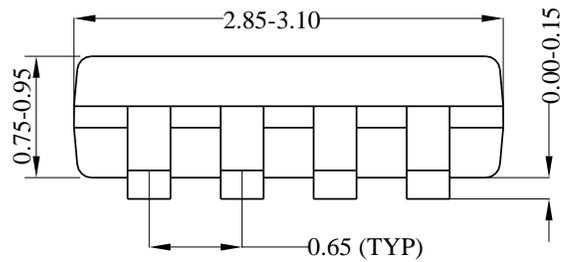
**Recommended pad layout**



**Top view**



**Side view A**

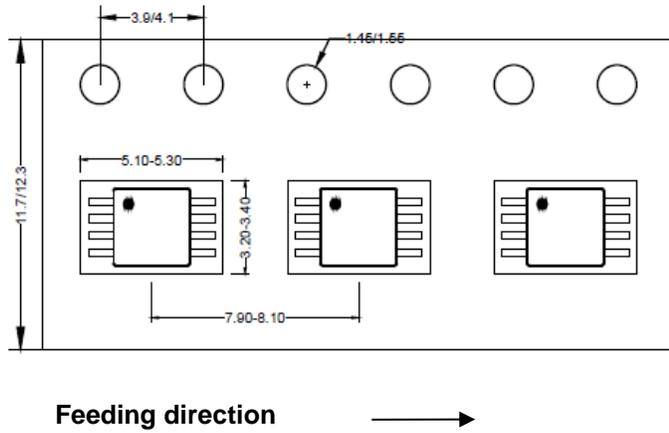


**Side view B**

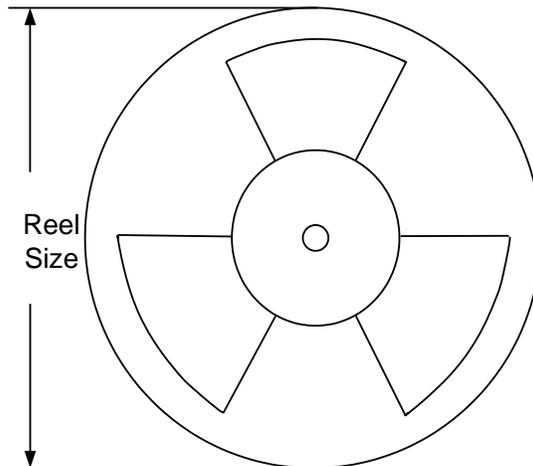
**Note:** All dimensions are in millimeters and exclude mold flash and metal burr.

# Taping and Reel Specification

## Package orientation



## Carrier tape and reel specification for packages



Package types	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Trailer length(mm)	Leader length (mm)	Qty per reel
MSOP8	12	8	13"	400	400	3000

Others: N/A

## Revision History

The revision history provided is for informational purpose only and is believed to be accurate, however, not warranted. Please make sure that you have the latest revision.

<b>Date</b>	<b>Revision</b>	<b>Change</b>
Sept.6, 2022	Revision 0.9	Initial Release.
Sept.6, 2023	Revision 1.0	Production Release.

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